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Quadruple Comparators

REJ03D0675-0300 Rev.3.00 Mar 10, 2006

Description

The HA17339A series products are comparators designed for general purpose, especially for power control systems.

These ICs operate from a single power-supply voltage over a wide range of voltages, and feature a reduced powersupply current since the supply current is independent of the supply voltage.

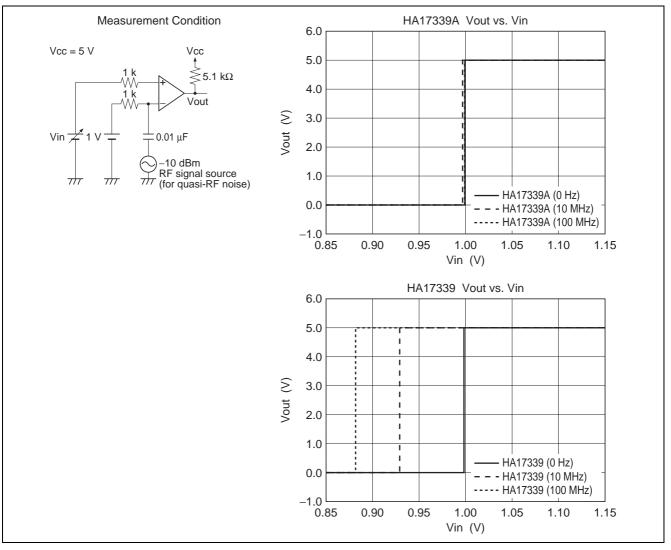
These comparators have the merit which ground is included in the common-mode input voltage range at a single-voltage power supply operation. These products have a wide range of applications, including limit comparators, simple A/D converters, pulse/square-wave/time delay generators, wide range VCO circuits, MOS clock timers, multivibrators, and high-voltage logic gates.

Features

- Wide power-supply voltage range : 2 to 36 V
- Very low supply current : 0.8 mA Typ.
- Low input bias current : 25 nA Typ.
- Low input offset current : 5 nA Typ.
- Low input offset voltage : 2 mV Typ.
- The common-mode input voltage range includes ground
- Output voltages compatible with CMOS logic systems



• Low electro-magnetic susceptibility

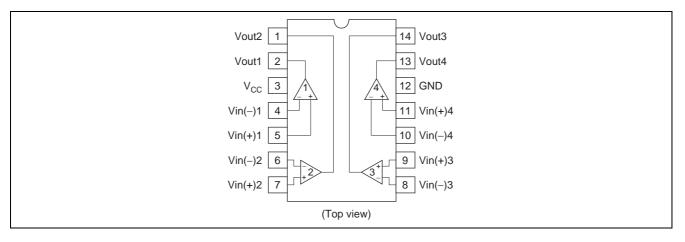


Ordering Information

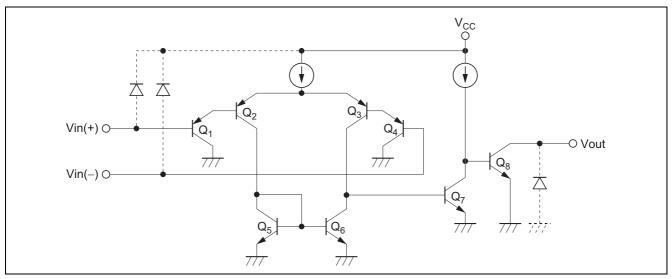
Type No.	Application	Package Name	Package Code
HA17339A	Commercial use	DIP-14 pin	PRDP0014AB-B
HA17339AF		SOP-14 pin (JEITA)	PRSP0014DF-B
HA17339ARP		SOP-14 pin (JEDEC)	PRSP0014DE-A
HA17339AT		TSSOP-14 pin	PTSP0014JA-B



Pin Arrangement



Circuit Structure (1/4)



Note: If Input/Output terminals voltage over the absolute maximum ratings, there is possibility of mis-operation, characteristics deterioration and destruction, because of the current's flowing to parasitic diode in IC. The Input/Output terminals are recommended to be protected with the clamp circuit which using the diode with low forward voltage (like schottky barrier diode) when there is a possibility for the Input/Output terminals voltage exceeds the absolute maximum ratings.



Absolute Maximum Ratings

				$(1a = 25^{\circ}C)$
ltem		Symbol	Ratings	Unit
Power supply voltage		V _{CC}	36	V
Differential input voltage		Vin(diff)	±V _{CC}	V
Input voltage		Vin	–0.3 to +V _{CC}	V
Output pin voltage		Vout	–0.3 to +36	V
Output current		lout *1	20	mA
Allowable power dissipation	DIP	PT	625 * ²	mW
	SOP		625 * ³	
	TSSOP		400 *4	
Operating temperature		Topr	-40 to +85	°C
Storage temperature		Tstg	–55 to +125	°C

Notes: 1. These products can be destroyed if the output and V_{CC} are shorted together. The maximum output current is the allowable value for continuous operation.

These are the allowable values up to $Ta = 50^{\circ}C$. Derate by 8.3 mW/°C above that temperature. 3. HA17339AF/ARP:

When it is mounted on glass epoxy board of 40 mm \times 40 mm \times 1.6 mmt with 10% wiring density, value at Ta \leq 25°C. If Ta > 25°C, derated by 6.25 mW/°C.

When it is mounted on glass epoxy board of 40 mm \times 40 mm \times 1.6 mmt with 30% wiring density. If Ta > 32°C, derated by 6.70 mW/°C.

4. HA17339AT:

These are the allowable values up to Ta = 25° C. Derate by 4 mW/°C above that temperature.

Electrical Characteristics

 $(V_{CC} = 5 V, Ta = 25^{\circ}C)$

 $(T_{0} - 25^{\circ}C)$

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Input offset voltage	V _{IO}	_	2	7	mV	Output switching point:
						when $V_0 = 1.4V$, $R_S = 0\Omega$
Input offset current	I _{IO}		5	50	nA	$ I_{IN(+)} - I_{IN(-)} $
Input bias current	I _{IB}	_	25	250	nA	I _{IN(+)} or I _{IN(-)}
Common-mode input voltage *1	V _{CM}	0		V _{CC} -1.5	V	
Supply current	Icc	_	0.8	2	mA	R _L = ∞
Voltage Gain * ³	Av	_	(200)	_	V/mV	$R_L = 15k\Omega$
Response time * ^{2,3}	t _R	_	(1.3)	_	μs	$V_{RL} = 5V, R_L = 5.1k\Omega$
Output sink current	I _{O(sink)}	6	16	_	mA	$V_{IN(-)} = 1V, V_{IN(+)} = 0, V_O \le 1.5V$
Output saturation voltage	V _{O(sat)}	_	200	400	mV	$V_{IN(-)} = 1V, V_{IN(+)} = 0, Iosink = 3mA$
Output leakage current *3	I _{LO}		(0.1)		nA	$V_{IN(+)} = 1V, V_{IN(-)} = 0, V_O = 5V$

Notes: 1. Voltages more negative than –0.3 V are not allowed for the common-mode input voltage or for either one of the input signal voltages.

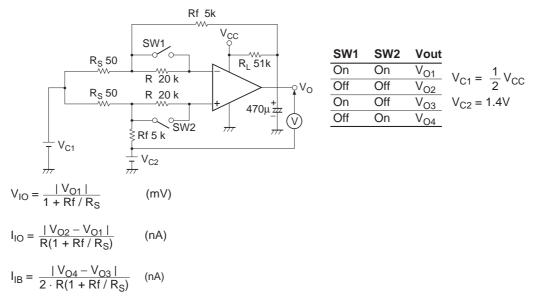
2. The stipulated response time is the value for a 100 mV input step voltage that has a 5 mV overdrive.

3. Design spec.

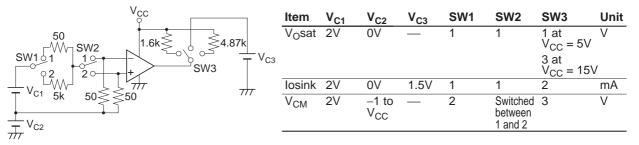
^{2.} HA17339A:

Test Circuits

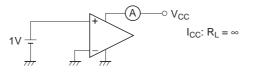
1. Input offset voltage (V_{IO}), input offset current (I_{IO}), and Input bias current (I_{IB}) test circuit



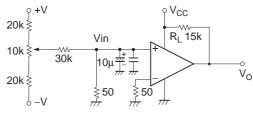
2. Output saturation voltage (V_O sat) output sink current (Iosink), and common-mode input voltage (V_{CM}) test circuit



3. Supply current (I_{CC}) test circuit

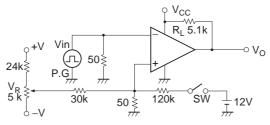


4. Voltage gain (A_V) test circuit ($R_L = 15k\Omega$)



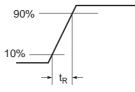
$$A_{V} = 20 \log \frac{V_{O1} - V_{O2}}{V_{IN1} - V_{IN2}} \quad (dB)$$

5. Response time (t_R) test circuit



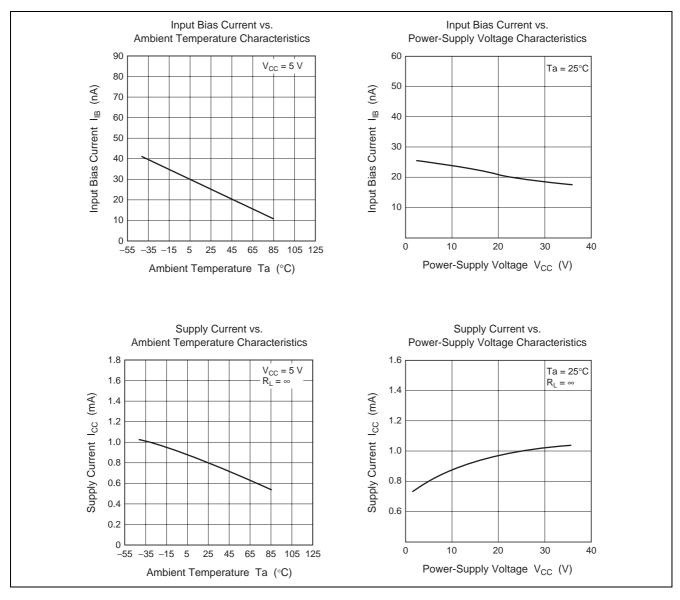
 $t_R : R_L = 5.1 k \Omega,$ a 100mV input step voltage that has a 5mV overdrive

- With V_{IN} not applied, set the switch SW to the off position and adjust V_R so that V_O is in the vicinity of 1.4V.
- Apply V_{IN} and turn the switch SW on.

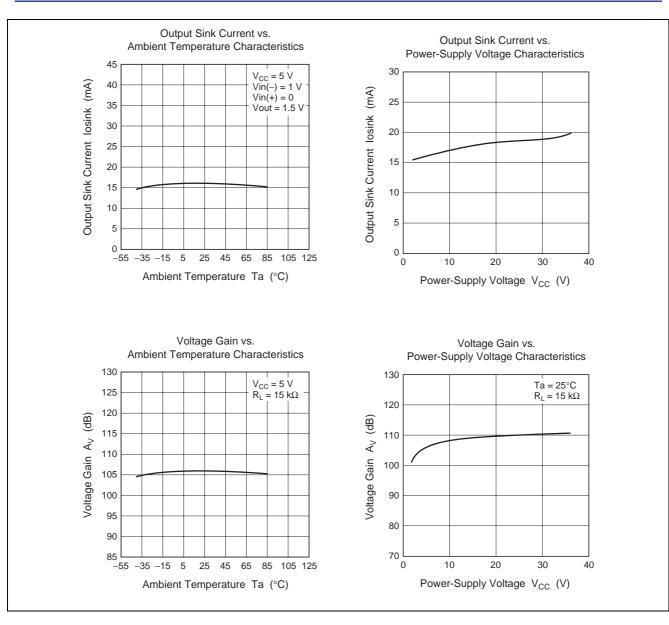




Characteristic Curves









HA17339A Application Examples

The HA17339A houses four independent comparators in a single package, and operates over a wide voltage range at low power from a single-voltage power supply. Since the common-mode input voltage range starts at the ground potential, the HA17339A is particularly suited for single-voltage power supply applications. This section presents several sample HA17339A applications.

HA17339A Application Notes

1. Square-Wave Oscillator

The circuit shown in figure one has the same structure as a single-voltage power supply astable multivibrator. Figure 2 shows the waveforms generated by this circuit.

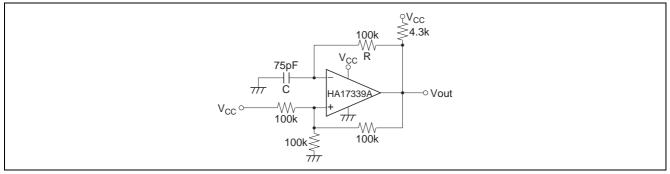


Figure 1 Square-Wave Oscillator

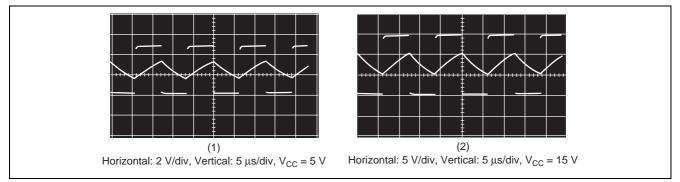


Figure 2 Operating Waveforms



2. Pulse Generator

The charge and discharge circuits in the circuit from figure 1 are separated by diodes in this circuit. (See figure 3.) This allows the pulse width and the duty cycle to be set independently. Figure 4 shows the waveforms generated by this circuit.

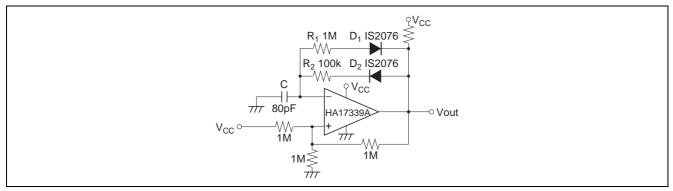


Figure 3 Pulse Generator

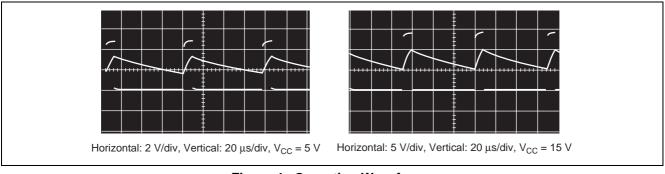


Figure 4 Operating Waveforms

3. Voltage Controlled Oscillator

In the circuit in figure 5, comparator A_1 operates as an integrator, A_2 operates as a comparator with hysteresis, and A_3 operates as the switch that controls the oscillator frequency. If the output Vout1 is at the low level, the A_3 output will go to the low level and the A1 inverting input will become a lower level than the A1 noninverting input. The A1 output will integrate this state and its output will increase towards the high level. When the output of the integrator A_1 exceeds the level on the comparator A_2 inverting input, A_2 inverts to the high level and both the output Vout1 and the A_3 output go to the high level. This causes the integrator to integrate a negative state, resulting in its output decreasing towards the low level. Then, when the A_1 output level becomes lower than the level on the A_2 noninverting input, the output Vout1 is once again inverted to the low level. This operation generates a square wave on Vout1 and a triangular wave on Vout2.

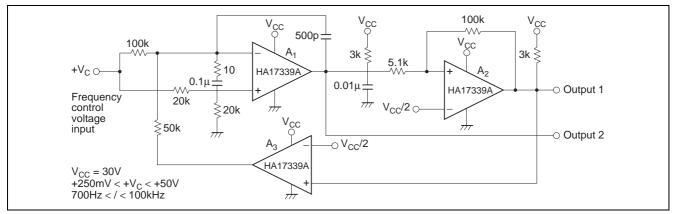


Figure 5 Voltage Controlled Oscillator



4. Basic Comparator

The circuit shown in figure 6 is a basic comparator. When the input voltage V_{IN} exceeds the reference voltage V_{REF} , the output goes to the high level.

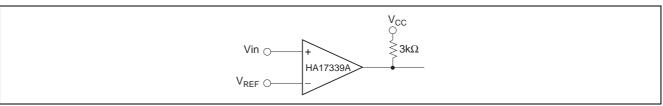


Figure 6 Basic Comparator

5. Noninverting Comparator (with Hysteresis)

Assuming $+V_{IN}$ is 0V, when V_{REF} is applied to the inverting input, the output will go to the low level (approximately 0V). If the voltage applied to $+V_{IN}$ is gradually increased, the output will go high when the value of the noninverting input, $+V_{IN} \times R_2/(R_1 + R_2)$, exceeds $+V_{REF}$. Next, if $+V_{IN}$ is gradually lowered, Vout will be inverted to the low level once again when the value of the noninverting input, $(Vout - V_{IN}) \times R_1/(R_1 + R_2)$, becomes lower than V_{REF} . With the circuit constants shown in figure 7, assuming $V_{CC} = 15V$ and $+V_{REF} = 6V$, the following formula can be derived, i.e. $+V_{IN} \times 10M/(5.1M + 10M) > 6V$, and Vout will invert from low to high when $+V_{IN}$ is > 9.06V.

$$(Vout - V_{IN}) \times \frac{R_1}{R_1 + R_2} + V_{IN} < 6V$$
(Assuming Vout = 15V)

When $+V_{IN}$ is lowered, the output will invert from high to low when $+V_{IN} < 1.41V$. Therefore this circuit has a hysteresis of 7.65V. Figure 8 shows the input characteristics.

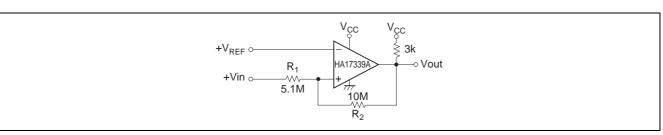


Figure 7 Noninverting Comparator

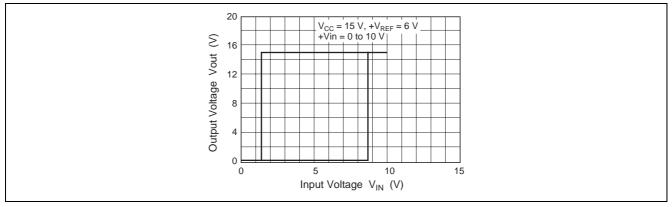


Figure 8 Noninverting Comparator I/O Transfer Characteristics



6. Inverting Comparator (with Hysteresis)

In this circuit, the output Vout inverts from high to low when $+V_{IN} > (V_{CC} + Vout)/3$. Similarly, the output Vout inverts from low to high when $+V_{IN} < V_{CC}/3$. With the circuit constants shown in figure 9, assuming $V_{CC} = 15V$ and Vout = 15V, this circuit will have a 5V hysteresis. Figure 10 shows the I/O characteristics for the circuit in figure 9.

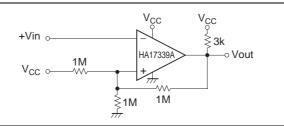


Figure 9 Inverting Comparator

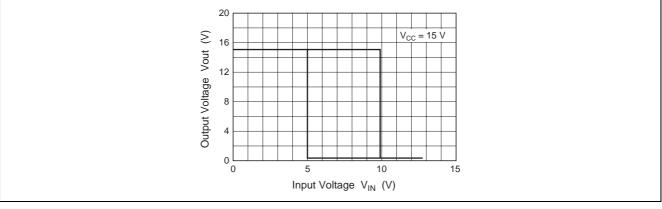


Figure 10 Inverting Comparator I/O Transfer Characteristics

7. Zero-Cross Detector (Single-Voltage Power Supply)

In this circuit, the noninverting input will essentially beheld at the potential determined by dividing V_{CC} with $100k\Omega$ and $10k\Omega$ resistors. When V_{IN} is 0V or higher, the output will be low, and when V_{IN} is negative, Vout will invert to the high level. (See figure 11.)

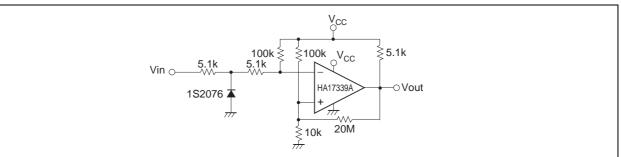
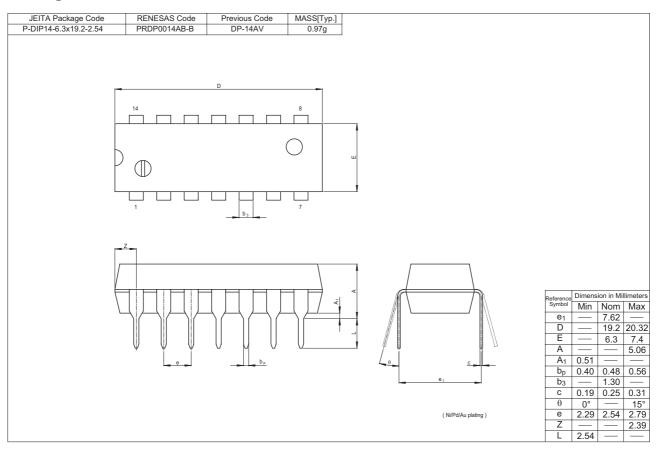
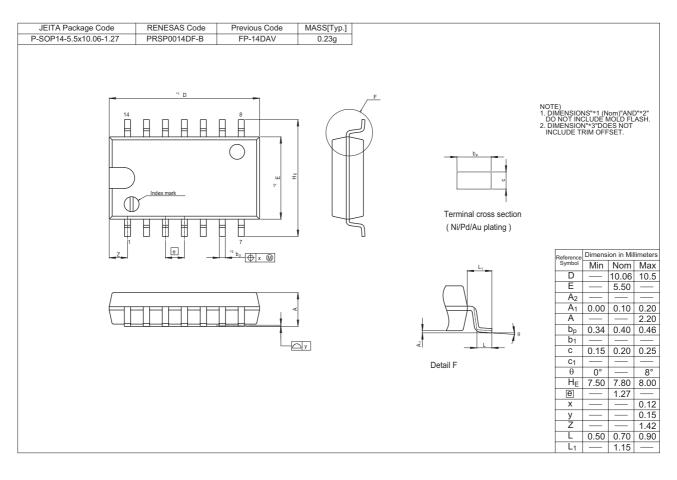


Figure 11 Zero-Cross Detector



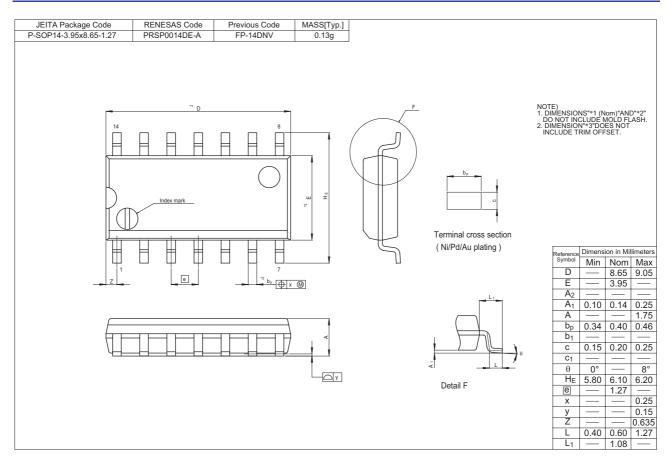
Package Dimensions

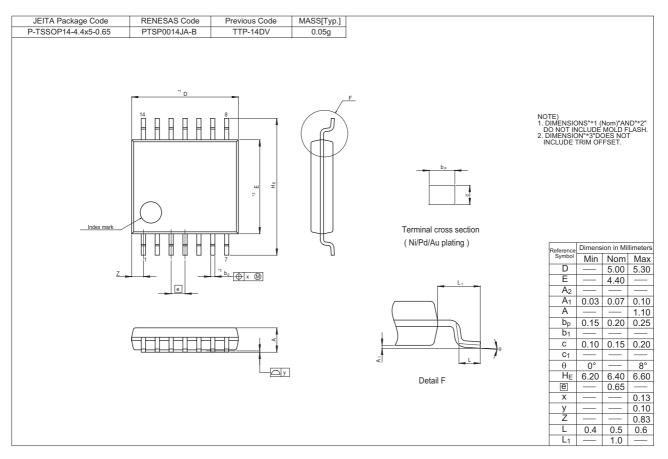




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