

Temperature Gradient SLG47011

# Abstract

This application note describes how to use the AnalogPAK SLG47011 to design a temperature gradient application. This application note comes complete with a design file that can be found in the Reference section.

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## 1. Terms and Definitions

Multichannel DCMPMultichannel Digital ComparatorPGAProgrammable Gain Amplifier

## 2. References

For related documents and software, please visit:

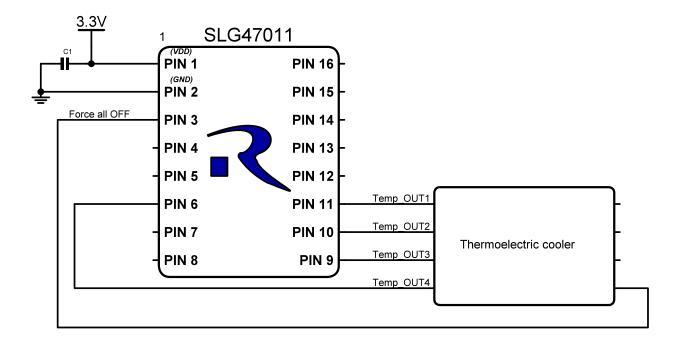
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Download our free Go Configure Software Hub [1] to open the design file [2] and view the proposed circuit design. Use the AnalogPAK development tools to freeze the design into your own customized IC in a matter of minutes. Renesas Electronics provides a complete library of application notes [3] featuring design examples, as well as explanations of features and blocks within the Renesas IC.

- [1] Go Configure Software Hub, Software Download and User Guide, Renesas Electronics
- [2] <u>AN-CM-402 Temperature Gradient.aap</u>, AnalogPAK Design File, Renesas Electronics
- [3] Application Notes, GreenPAK Application Notes Webpage, Renesas Electronics

## 3. Introduction

The input data in a Temperature Gradient application consists of a temperature value from an internal temperature sensor which can be used in applications where the rate of change of temperature needs to be detected (e.g., refrigerators).





# 4. AnalogPAK Design

In this design, the temperature gradient (rate of change) with four levels of detection provided by a digital comparator, takes the current temperature value and stores it in Buffer0. The previous temperature value is stored in Buffer1 and the difference between the current and previous values are compared by the Multichannel DCMP. Adjusting the hysteresis value can set the detection speed of the rise or drop in temperature and the four outputs are used to find the gradient.

On the power controller, Retention Mode can be performed during non-measurement periods through mode switching. In this mode, the quiescent current is less than 5  $\mu$ A. This design maintains data storage and saves power.

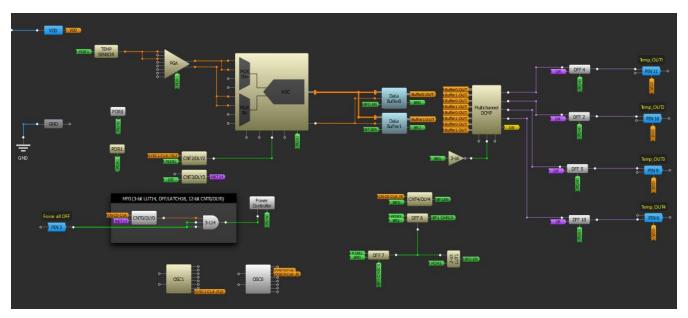


Figure 2. GreenPAK Designer Schematic of the Temperature Gradient application

### 4.1 Temperature Sensor

The SLG47011 has an Analog Temperature sensor (TS) with an output voltage linearly proportional to the Centigrade temperature. The TS output is selected as a source to the ADC or ACMP positive input. The TS is rated to operate over a -40 °C to 85 °C temperature range. The error in the entire temperature range does not exceed  $\pm 1.5$  % and the TS output voltage variation over V<sub>DD</sub> at constant temperature is less than  $\pm 1.5$  %.

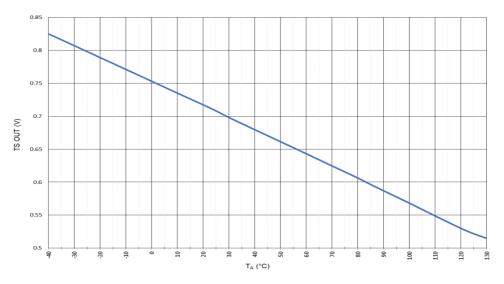


Figure 3. TS Output vs. Temperature,  $V_{\text{DD}}$  = 1.71 V to 3.6 V

## 4.2 PGA Configuration

In temperature sensing applications, the PGA of the SLG47011 is configured to operate in mode 6. The value from the internal temperature is converted into a voltage and the signal is sent to the ADC through the buffer.

The PGA channel settings are shown in Figure 4.

roperties		
	PGA	
Out+ to PIN 13 (GPIO9)	Disable <b>*</b>	
Out- to PIN 14 (GPIO10) PGA/ADC	Disable <b>*</b>	Mode 6: Single-ended Buffer
manual mode enable:	Disable •	
Manual channel selection:	Channel 0 💌	
C	hannel 0	
Input mode:	Single ended inpu 🔻	
Mode:	(6) Buffer 💌	
Gain:	1x *	
IN+ source:	TEMP SENSOR	
IN- source:	AGND	
C	hannel 1	
Input mode:	Single ended inpu 🔻	GPIO ADC
Mode:	(6) Buffer 👻	ADC
Gain:	1x ···	
IN+ source:	TEMP SENSOR *	=
IN- source:	AGND -	

Figure 4. PGA configuration using channel 0 and channel 1 to compare the previous and current temperature.

### 4.3 ADC Configuration

The ADC is configured to use OSC1 with the 20 MHz clock selected. Please note that the Clock divider is set to the /16 divider (31.25 kps). Since the temperature is detected in sequence, this setting will allow each channel to sense the delay.

Properties	I
	ADC
Clock selection:	OSC1 -
Vref selection:	1.62V internal Vret 🔻
AVDD divider:	(1/8)AVDD -
Resolution:	14-bit 👻
Sample per channel:	1
Channel 0 system calibration:	Disable •
Channel 2 system calibration:	Disable •
Clock divider:	/16 divider 💌
Sampling rate (single channel):	31.250 ksps <u>Formul</u>
Delay between channels:	5
Delay between channels predivider:	8 -
Delay:	32 us
Data aligment:	MSB 🔻

Figure 5. ADC Configuration

### 4.4 Data Buffer0/1 Configuration

Mode: Moving Average

Length: 1 word

Buffer ready: 1

Properties		Properties		×	
Data Buffer0			Data Buffer1		
Mode:	Moving Average	-	Mode:	Moving Average	•
Length:	1 word	-	Length:	1 word	-
Initial data:	0000h	•	Initial data:	0000h	-
Input source:	ADC	•	Input source:	ADC	-
Load source:	ADC ready 0	•	Load source:	ADC ready 1	-
Load en sync:	No sync	•	Load en sync:	No sync	-
OUT source:	Data	•	OUT source:	Data	-
Buffer ready:	1	-	Buffer ready:	1	•
	Apply Apply			Apply	

Figure 6. Data Buffer0 and Data Buffer1 Configuration

### 4.5 Multichannel DCMP Configuration

Mode: Sequence conversion Mode

Sync enable: Async

Eanble Source: Matrix out

Compare selection: Greater than

IN+ source: Buffer0 data

IN- source: Buffer1 data

Hysteresis value: Adjust according to gradient requirement

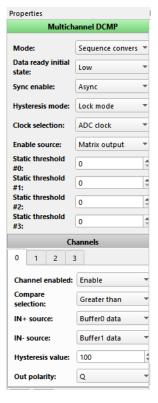


Figure 7. Multichannel DCMP Configuration

### 4.5.1. Adjusting the Hysteresis value:

The hysteresis value is calculated based on the slope of temperature change versus time according to the following formulas.

In section 4.1, it is mentioned that the following values are obtained based on the temperature versus voltage output of the SLG47011's internal temperature sensor:

$$V_{TS \ OUT} = 0.0019 \left[ \frac{V}{\circ c} \right]$$

In temperature gradient application designs, two design parameters must be known. The first is the temperature change per second and the section is the detection interval.

Temperature change per second =  $\frac{\Delta \text{Temperature}}{\Delta \text{Time}} \left[ \frac{\Delta T}{Sec} \right]$ 

Next, the temperature change per second is multiplied by the detection time and then further multiplied by the voltage value represented by  $^{\circ}$  C per degree to obtain the Hysteresis voltage change HYS ( $\alpha$ ).

 $HYS(\alpha) = Temperature change per second \times Detection interval \times V_{TS OUT}$ 

Where the detection interval is the interval time between front and rear temperature detection.

Lastly, we divide the obtained HYS ( $\alpha$ ) by the ADC reference voltage value Vref (1.62 V) and multiply it by the ADC resolution (14-bit = 16383) to get the design threshold Hysteresis value.

$$Hysteresis Value = \frac{HYS(\alpha)}{Vref(1.62V)} \times 16383 (ADC resolution, 14bits)$$

### 4.6 Operating Instructions

The design operates in three stages:

First, the data from the TEMP SENSOR is written to Buffer0, and after a specified delay, which is configured using a One shot (CNT/DLY4), the data from the TEMP SENSOR is written to Buffer1. The data from Buffer0 and Buffer1 are compared by the Multichannel DCMP during this cycle. Depending on the difference between the data in Data Buffer0 and Data Buffer1, we can determine the rate of temperature change.

The following tests use a range of voltage values which are used as a temperature change simulation, and four groups of experiments were designed to verify the actual application examples.

### Test conditions:

Temperature range	$25^{\circ}$ C – $40^{\circ}$ C	$25^{\circ}$ C – $60^{\circ}$ C	$25^{\circ}$ C – $80^{\circ}$ C	$25^\circ~$ C – $100^\circ~$ C		
Voltage range	705 mV – 667 mV	705 mV – 638 mV	705 mV – 600 mV	705 mV – 562 mV		
Time	6 seconds					
Hysteresis value	Temp_OUT1: 100 Temp_OUT2: 300 Temp_OUT3: 500 Temp_OUT4: 600					

# 5. Functional Waveforms

Channel 1 (yellow/top line) – TS Output D0 – Temp\_OUT1 D1 – Temp\_OUT2 D2 – Temp\_OUT3

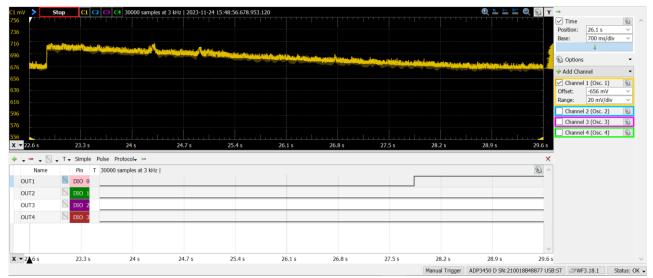


Figure 8. Temperature rise 25 °C – 40 °C

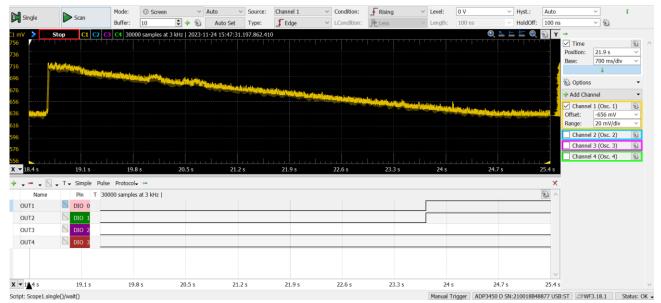
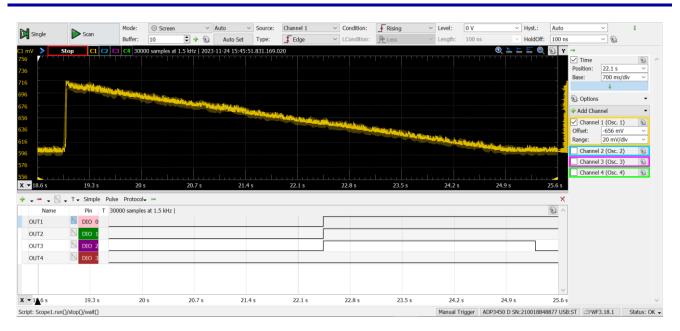


Figure 9. Temperature rise 25 °C - 60 °C

### **Temperature Gradient**





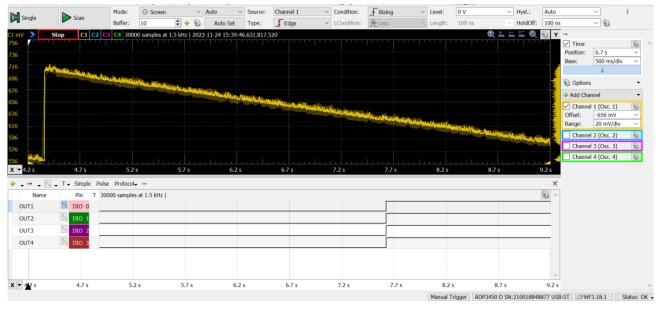


Figure 11. Temperature rise 25 °C – 100 °C

# 6. Conclusion

From the above four test results, the yellow line is the voltage value corresponding to the internal temperature sensing when the temperature rises, the test time is all 6 seconds, the range of temperature rise affects the slope. The first test temperature of  $25^{\circ}$  C to  $40^{\circ}$  C has a relatively slow rising slope and can only trigger the output of OUT1. The fourth test temperature of  $25^{\circ}$  C to  $100^{\circ}$  C, the rising slope is steeper and Temp\_OUT1 - Temp\_OUT4 are all triggered.

Therefore, it can be known from the 4 test results that the number of OUT will trigger different outputs as the temperature rise slope becomes slower and steeper, and by adjusting the Hysteresis value in the Multichannel DCMP parameter, different slope detection can be set.

# 7. Revision History

Revision	Date	Description
1.00	November 5, 2024	Initial release

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