

## General Description

The ICS840N011I is an LVCMOS/LVTTL clock synthesizer designed for Fibre Channel applications. The device generates a 106.25MHz clock signal from a 26.5625MHz crystal or a 100MHz clock signal from a 25MHz crystal with excellent phase jitter performance. The device uses IDT's fourth generation FemtoClock® NG technology for an optimum of high clock frequency, low phase noise performance and low power consumption and high power supply noise rejection. The device supports 2.5V or 3.3V voltage supply and is packaged in a small, lead-free (RoHS 6) 8-lead TSSOP package. The extended temperature range supports wireless infrastructure, telecommunication and networking end equipment requirements.

### Frequency Table

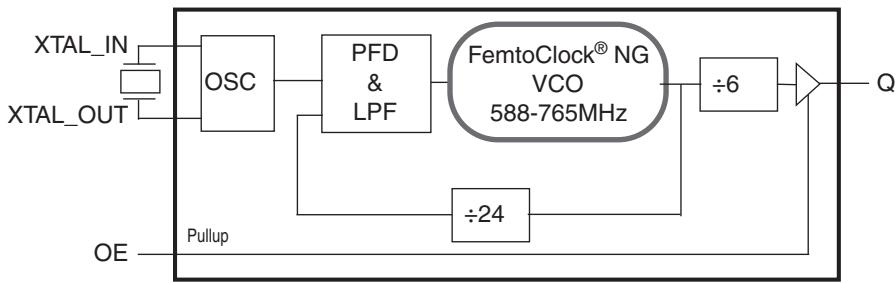
$f_{XTAL}$ (MHz)	Output Frequency (MHz)
25	100
26.5625	106.25
30.72	122.88
31.25	125

### Function Table

Input	Output Enable
OE	
0	Output Q is disabled in high-impedance state
1 (default)	Output Q is enabled.

NOTE: OE is an asynchronous control.

## Block Diagram



## Features

- Fourth generation FemtoClock® NG technology
- 106.25MHz output clock synthesized from a 26.5625MHz fundamental mode crystal
- One 2.5V or 3.3V LVCMOS/LVTTL clock output
- Crystal interface designed for a 12pF parallel resonant crystal
- RMS phase jitter @ 100MHz, using a 25MHz crystal (637kHz - 10MHz): 0.185ps (maximum)
- LVCMOS/LVTTL interface level for the output enable input
- Full 2.5V or 3.3V supply voltage
- Lead-free (RoHS 6) packaging
- 40°C to 85°C ambient operating temperature

## Pin Assignment

VDDA	1	8	VDD
OE	2	7	Q
XTAL_OUT	3	6	GND
XTAL_IN	4	5	DNU

ICS840N011I  
8-lead TSSOP  
4.40mm x 3.0mm x 0.925mm  
package body  
G Package  
Top View

## Pin Descriptions and Characteristics

**Table 1. Pin Descriptions**

Number	Name	Type		Description
1	$V_{DDA}$	Power		Analog power supply.
2	OE	Input	Pullup	Output enable pin. LVCMS interface levels.
3, 4	XTAL_OUT, XTAL_IN	Input		Crystal oscillator interface. XTAL_IN is the input, XTAL_OUT is the output.
5	DNU			Do not use. Do not connect.
6	GND	Power		Power supply ground.
7	Q	Output		Single-ended clock output. LVCMS/LVTTL interface levels.
8	$V_{DD}$	Power		Core supply pin.

NOTE: *Pullup* refers to an internal input resistor. See Table 2, *Pin Characteristics*, for typical values.

**Table 2. Pin Characteristics**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$C_{IN}$	Input Capacitance	OE		3.5		pF
$C_{PD}$	Power Dissipation Capacitance	$V_{DD} = 3.465V$		11		pF
		$V_{DD} = 2.625V$		9		pF
$R_{PULLUP}$	Input Pullup Resistor			51		k $\Omega$
$R_{OUT}$	Output Impedance	$V_{DD} = 3.3V$		15		$\Omega$
		$V_{DD} = 2.5V$		19		$\Omega$

## Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, $V_{DD}$	3.63V
Inputs, $V_I$ XTAL_IN Other Inputs	0V to 2V -0.5V to $V_{DD} + 0.5V$
Outputs, $V_O$	-0.5V to $V_{DD} + 0.5V$
Package Thermal Impedance, $\theta_{JA}$	117°C/W (0 mps)
Storage Temperature, $T_{STG}$	-65°C to 150°C

## DC Electrical Characteristics

Table 3A. Power Supply DC Characteristics,  $V_{DD} = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$ ,  $T_A = -40^\circ C$  to  $85^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Core Supply Voltage		2.375	3.3	3.465	V
$V_{DDA}$	Analog Supply Voltage		$V_{DD} - 0.18$	3.3	$V_{DD}$	V
$V_{DDA}$	Analog Supply Voltage		$V_{DD} - 0.18$	2.5	$V_{DD}$	V
$I_{DDA}$	Analog Supply Current				18	mA
$I_{DD}$	Power Supply Current				67	mA

Table 3B. LVC MOS/LV TTL DC Characteristics,  $V_{DD} = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$ ,  $T_A = -40^\circ C$  to  $85^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{IH}$	Input High Voltage	$V_{DD} = 3.3V$	2		$V_{DD} + 0.3$	V
		$V_{DD} = 2.5V$	1.7		$V_{DD} + 0.3$	V
$V_{IL}$	Input Low Voltage	$V_{DD} = 3.3V$	-0.3		0.8	V
		$V_{DD} = 2.5V$	-0.3		0.7	V
$I_{IH}$	Input High Current	OE	$V_{DD} = V_{IN} = 3.465V$ or $2.625V$		5	$\mu A$
$I_{IL}$	Input Low Current	OE	$V_{DD} = 3.465V$ $3.465V$ or $2.625V$ , $V_{IN} = 0V$	-150		$\mu A$
$V_{OH}$	Output High Voltage; NOTE 1	Q	$V_{DD} = 3.465V$	2.6		V
			$V_{DD} = 2.625V$	1.8		V
$V_{OL}$	Output Low Voltage; NOTE 1	Q	$V_{DD} = 3.465V$ or $2.625V$		0.5	V

NOTE 1: Output terminated with  $50\Omega$  to  $V_{DD} / 2$ . See Parameter Measurement Information Section, *LVC MOS Output Load Test Circuit Diagrams*.

**Table 4. Crystal Characteristics**

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation		Fundamental			
Frequency		24.50	26.5625	31.88	MHz
Equivalent Series Resistance (ESR)				80	$\Omega$
Shunt Capacitance				7	pF
Capacitive Load ( $C_L$ )			12		pF

## AC Characteristics

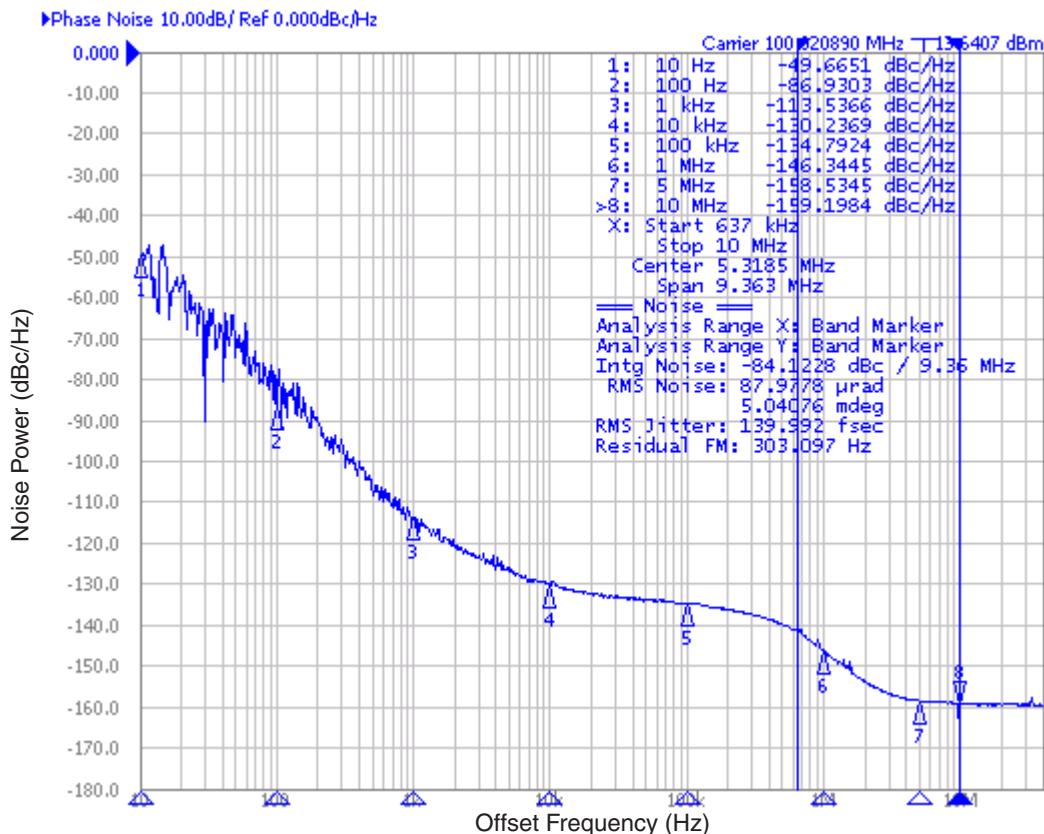
**Table 5. AC Characteristics,  $V_{DD} = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$ ,  $T_A = -40^\circ C$  to  $85^\circ C$** 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$f_{OUT}$	Output Frequency		98.00	106.25	127.52	MHz
$\text{f}_{\text{jit}}(\emptyset)$	RMS Phase Jitter (Random); NOTE 1	$f_{OUT} = 100\text{MHz}$ , 25MHz Crystal, Integration Range: 637kHz – 10MHz		0.140	0.185	ps
		$f_{OUT} = 106.25\text{MHz}$ , 26.5625MHz Crystal, Integration Range: 637kHz – 10MHz		0.139	0.177	ps
$\Phi_N$	Single-Side Band Noise Power	$f_{OUT} = 106.25\text{MHz}$ , Offset: 10Hz		-60.4		dBc/Hz
		$f_{OUT} = 106.25\text{MHz}$ , Offset: 100Hz		-87.4		dBc/Hz
		$f_{OUT} = 106.25\text{MHz}$ , Offset: 1kHz		-117.8		dBc/Hz
		$f_{OUT} = 106.25\text{MHz}$ , Offset: 10kHz		-130.7		dBc/Hz
		$f_{OUT} = 106.25\text{MHz}$ , Offset: 100kHz		-134.9		dBc/Hz
		$f_{OUT} = 106.25\text{MHz}$ , Offset: 1MHz		-145.6		dBc/Hz
		$f_{OUT} = 106.25\text{MHz}$ , Offset: 10MHz		-158.9		dBc/Hz
$t_R / t_F$	Output Rise/Fall Time	20% to 80%	200		600	ps
odc	Output Duty Cycle		48		52	%

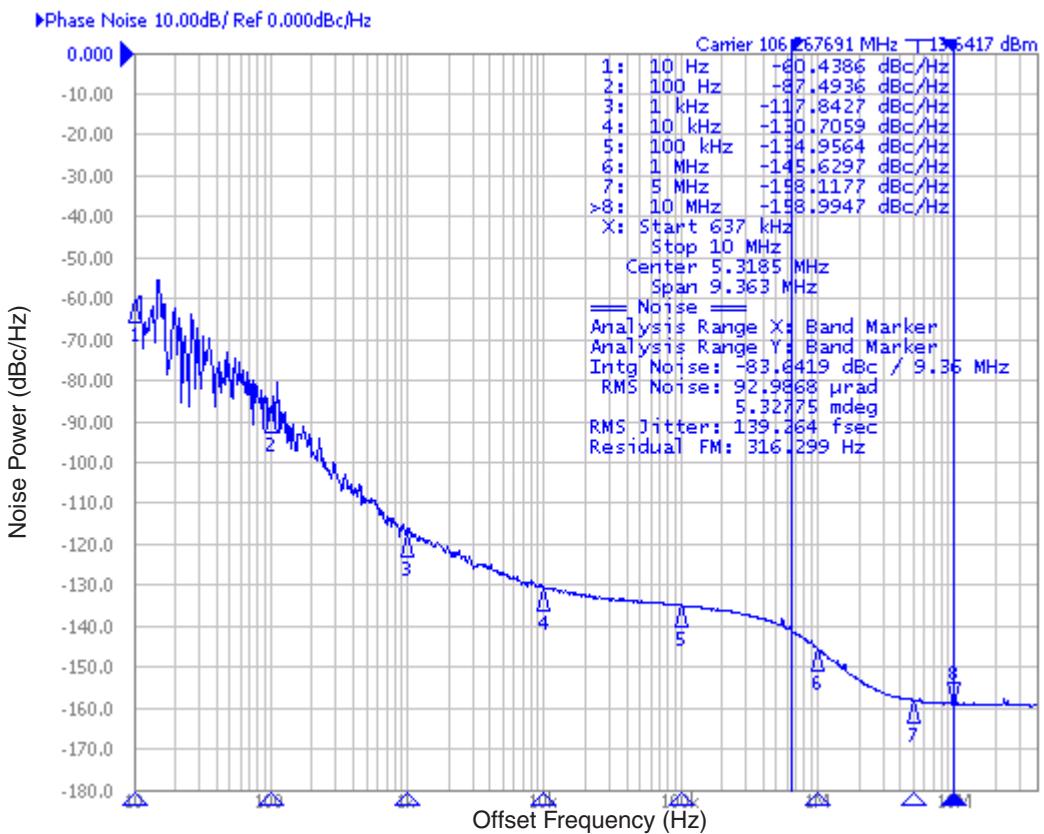
NOTE: Characterized with 25MHz, 26.5625MHz, 30.72MHz and 31.25MHz crystals.

NOTE 1: Please refer to the phase noise plots.

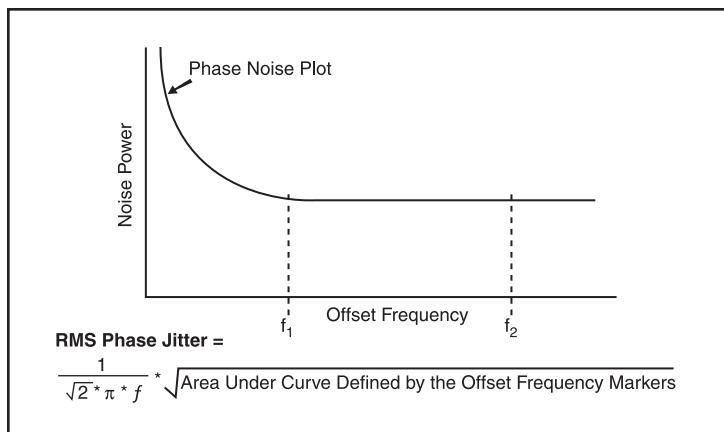
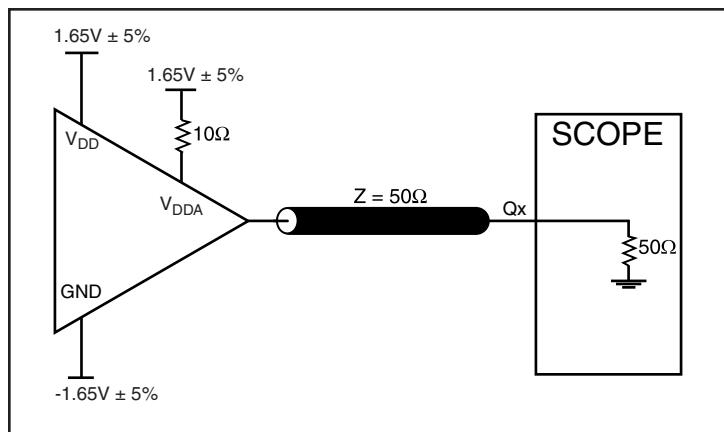
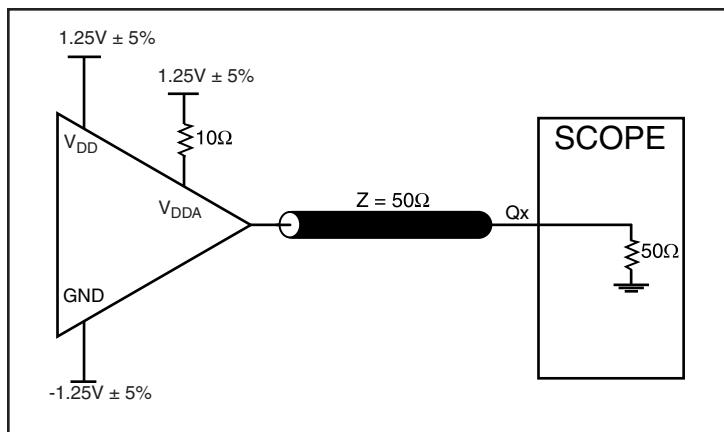
## Typical Phase Noise at 100MHz (637kHz - 10MHz)



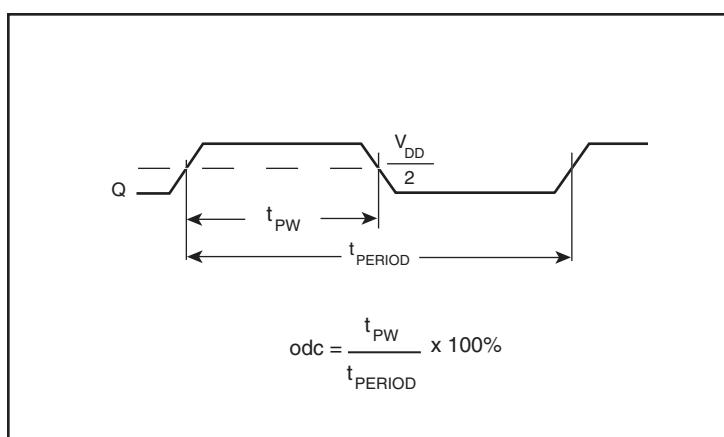
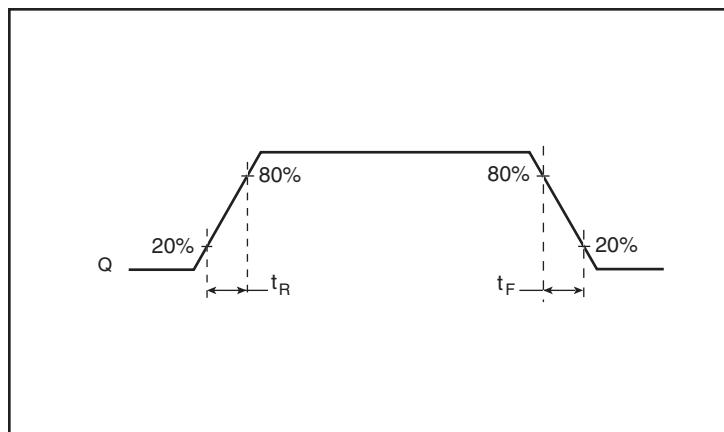
## Typical Phase Noise at 106.25MHz (637kHz - 10MHz)



## Parameter Measurement Information



RMS Phase Jitter



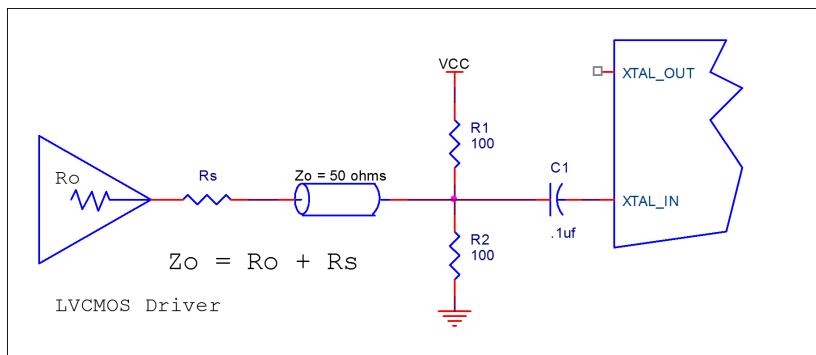
Output Duty Cycle/Pulse Width/Period

## Applications Information

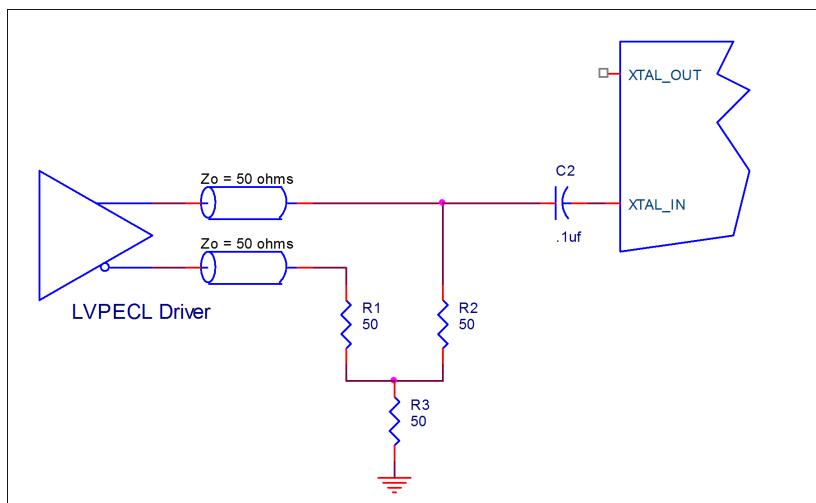
### Overdriving the XTAL Interface

The XTAL\_IN input can be overdriven by an LVC MOS driver or by one side of a differential driver through an AC coupling capacitor. The XTAL\_OUT pin can be left floating. The amplitude of the input signal should be between 500mV and 1.8V and the slew rate should not be less than 0.2V/ns. For 3.3V LVC MOS inputs, the amplitude must be reduced from full swing to at least half the swing in order to prevent signal interference with the power rail and to reduce internal noise. *Figure 1A* shows an example of the interface diagram for a high speed 3.3V LVC MOS driver. This configuration requires that the sum of the output impedance of the driver ( $Z_o$ ) and the series resistance ( $R_s$ ) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This

can be done in one of two ways. First,  $R_1$  and  $R_2$  in parallel should equal the transmission line impedance. For most  $50\Omega$  applications,  $R_1$  and  $R_2$  can be  $100\Omega$ . This can also be accomplished by removing  $R_1$  and changing  $R_2$  to  $50\Omega$ . The values of the resistors can be increased to reduce the loading for a slower and weaker LVC MOS driver. *Figure 1B* shows an example of the interface diagram for an LVPECL driver. This is a standard LVPECL termination with one side of the driver feeding the XTAL\_IN input. It is recommended that all components in the schematics be placed in the layout. Though some components might not be used, they can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a quartz crystal as the input.



**Figure 1A. General Diagram for LVC MOS Driver to XTAL Input Interface**



**Figure 1B. General Diagram for LVPECL Driver to XTAL Input Interface**

## Schematic Layout

Figure 2 shows an example ICS840N011I application schematic in which the device is operated at  $V_{DD} = V_{DDA} = +3.3V$ . The schematic example focuses on functional connections and is intended as an example only and may not represent the exact user configuration. Refer to the pin description and functional tables in the datasheet to ensure the logic control inputs are properly set. For example OE can be configured from an FPGA instead of set with pull up and pull down resistors as shown.

As with any high speed analog circuitry, the power supply pins are vulnerable to random noise, so to achieve optimum jitter performance isolation of the  $V_{DD}$  pin from power supply is required. In order to achieve the best possible filtering, it is recommended that the placement of the filter components be on the device side of the PCB as close to the power pins as possible. If space is limited, the  $0.1\mu F$

capacitor on the  $V_{DD}$  pin must be placed on the device side with direct return to the ground plane though vias. The remaining filter components can be on the opposite side of the PCB.

Power supply filter component recommendations are a general guideline to be used for reducing external noise from coupling into the devices. The filter performance is designed for a wide range of noise frequencies. This low-pass filter starts to attenuate noise at approximately 10kHz. If a specific frequency noise component is known, such as switching power supplies frequencies, it is recommended that component values be adjusted and if required, additional filtering be added. Additionally, good general design practices for power plane voltage stability suggests adding bulk capacitance in the local area of all devices.

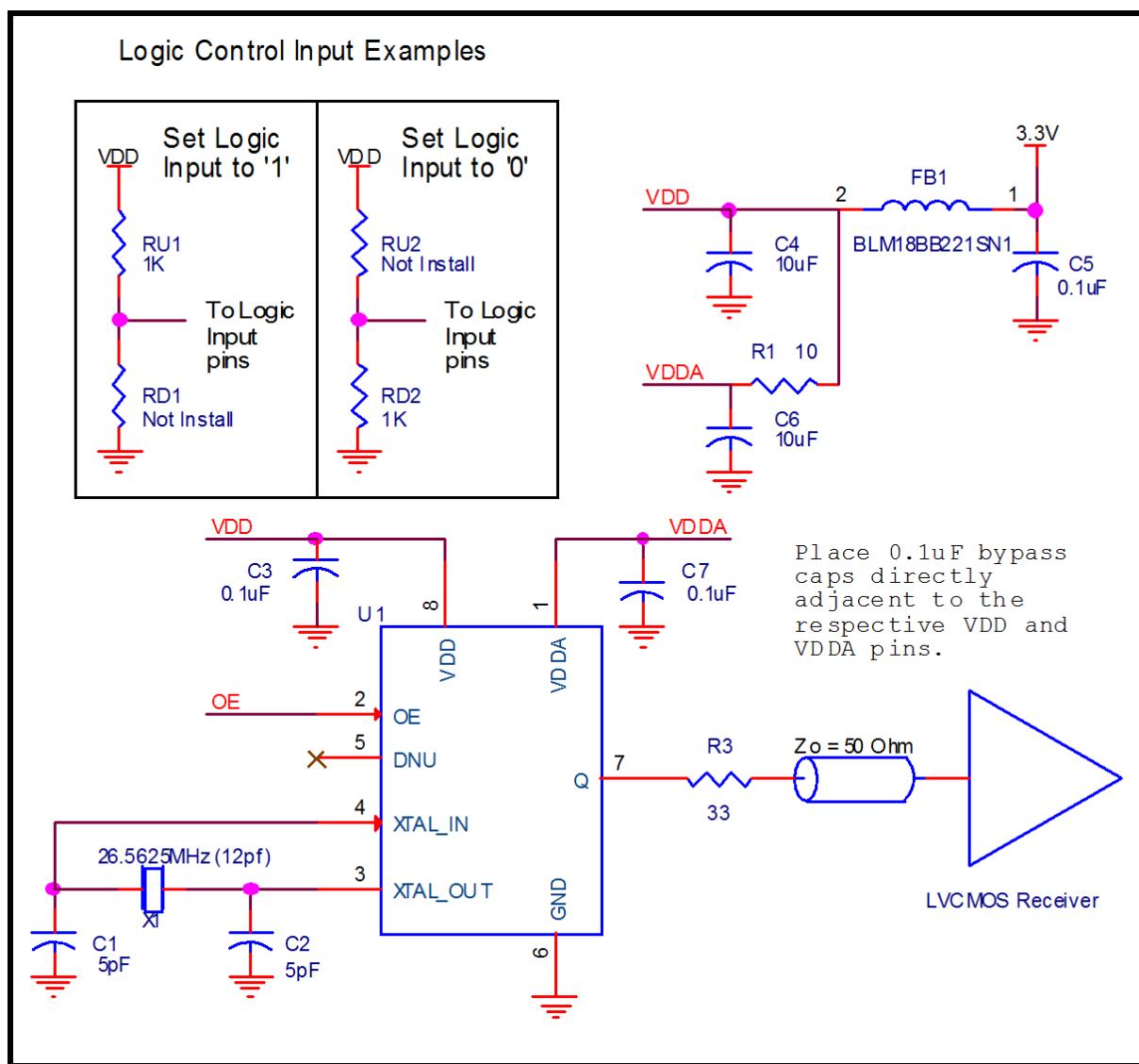


Figure 2. ICS840N011I Application Schematic

## Power Considerations

This section provides information on power dissipation and junction temperature for the ICS840N011I. Equations and example calculations are also provided.

### 1. Power Dissipation.

The total power dissipation for the ICS840N011I is the sum of the core power plus the analog power plus the power dissipated in the load(s). The following is the power dissipation for  $V_{DD} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

- Power (core)<sub>MAX</sub> =  $V_{DD\_MAX} * (I_{DD} + I_{DDA}) = 3.465V * (67mA + 18mA) = 294.53mW$
- Output Impedance  $R_{OUT}$  Current due to Loading  $50\Omega$  to  $V_{DD}/2$   

$$Output\ Current\ I_{OUT} = V_{DD\_MAX} / [2 * (50\Omega + R_{OUT})] = 3.465V / [2 * (50\Omega + 15\Omega)] = 26.7mA$$
- Power Dissipation on the  $R_{OUT}$  per LVC MOS output  

$$Power (R_{OUT}) = R_{OUT} * (I_{OUT})^2 = 15\Omega * (26.7mA)^2 = 10.7mW$$
- Total Power ( $R_{OUT}$ ) =  $10.7mW * 1 = 10.7mW$

#### Dynamic Power Dissipation at 125MHz

$$Power (125MHz) = C_{PD} * Frequency * (V_{DD})^2 = 11pF * 125MHz * (3.465V)^2 = 16.51mW$$

$$Total\ Power (125MHz) = 16.51mW * 1 = 16.51mW$$

#### Total Power Dissipation

- **Total Power**  

$$= Power (core)_{MAX} + Power (R_{OUT}) + Power (125MHz)$$
  

$$= 294.53mW + 10.7mW + 16.51mW$$
  

$$= 321.74mW$$

### 2. Junction Temperature.

Junction temperature,  $T_j$ , is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is  $125^\circ\text{C}$ . Limiting the internal transistor junction temperature,  $T_j$ , to  $125^\circ\text{C}$  ensures that the bond wire and bond pad temperature remains below  $125^\circ\text{C}$ .

The equation for  $T_j$  is as follows:  $T_j = \theta_{JA} * P_{d\_total} + T_A$

$T_j$  = Junction Temperature

$\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

$P_{d\_total}$  = Total Device Power Dissipation (example calculation is in section 1 above)

$T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is  $117^\circ\text{C}/\text{W}$  per Table 6 below.

Therefore,  $T_j$  for an ambient temperature of  $85^\circ\text{C}$  with all outputs switching is:

$$85^\circ\text{C} + 0.322W * 117^\circ\text{C}/\text{W} = 122.7^\circ\text{C}. This is below the limit of 125^\circ\text{C}.$$

This calculation is only an example.  $T_j$  will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

**Table 6. Thermal Resistance  $\theta_{JA}$  for 8 Lead TSSOP, Forced Convection**

$\theta_{JA}$ by Velocity	
Meters per Second	0
Multi-Layer PCB, JEDEC Standard Test Boards	$117^\circ\text{C}/\text{W}$

## Reliability Information

Table 7.  $\theta_{JA}$  vs. Air Flow Table for a 8-lead TSSOP

$\theta_{JA}$ vs. Air Flow	
Meters per Second	0
Multi-Layer PCB, JEDEC Standard Test Boards	117°C/W

## Transistor Count

The transistor count for ICS840N011I is: 24,811

## Package Outline and Package Dimensions

Package Outline - G Suffix for 8 Lead TSSOP

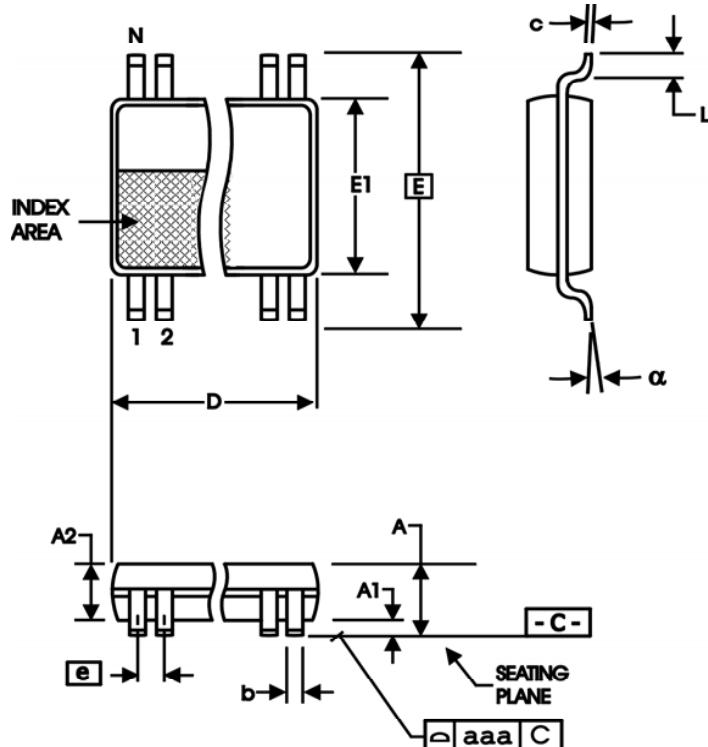


Table 8. Package Dimensions

All Dimensions in Millimeters		
Symbol	Minimum	Maximum
N		8
A		1.20
A1	0.5	0.15
A2	0.80	1.05
b	0.19	0.30
c	0.09	0.20
D	2.90	3.10
E	6.40 Basic	
E1	4.30	4.50
e	0.65 Basic	
L	0.45	0.75
$\alpha$	0°	8°
aaa		0.10

Reference Document: JEDEC Publication 95, MO-153

**Table 9. Ordering Information**

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
840N011BGILF	11BIL	Lead-Free, 8-lead TSSOP	Tube	-40°C to 85°C
840N011BGILFT	11BIL	Lead-Free, 8-lead TSSOP	Tape & Reel	-40°C to 85°C



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### Corporate Headquarters

TOYOSU FORESIA, 3-2-24 Toyosu,  
Koto-ku, Tokyo 135-0061, Japan  
[www.renesas.com](http://www.renesas.com)

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