

### General Description

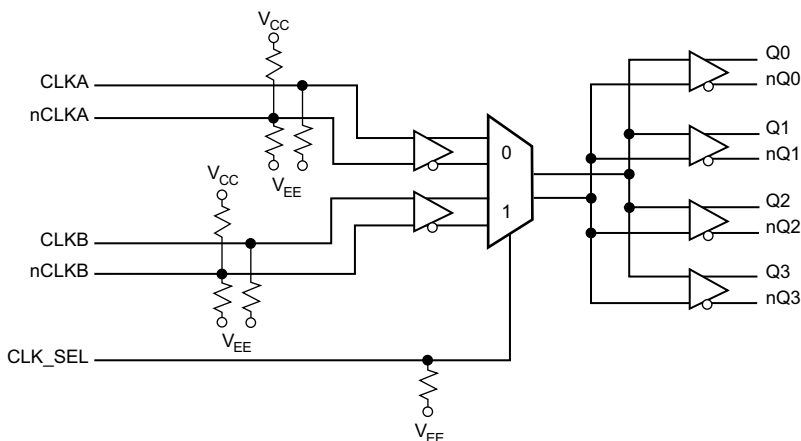
The IDT8T33FS314I is a low skew 1-to-4 Differential Fanout Buffer, designed with clock distribution in mind, accepting two clock sources into an input MUX. The MUX is controlled by a CLK\_SEL pin. This makes the IDT8T33FS314I very versatile, in that, it can operate as both a differential clock buffer as well as a signal-level translator and fanout buffer.

The device is designed on a SiGe process and can operate at frequencies in excess of 2.7GHz. This ensures negligible jitter introduction to the timing budget which makes it an ideal choice for distributing high frequency, high precision clocks across back planes and boards in communication systems. Internal temperature compensation guarantees consistent performance across various platforms.

### Features

- Four differential ECL/LVPECL level outputs
- One differential ECL/LVPECL or single-ended input (CLKA)  
One differential HSTL or single-ended input (CLKB)
- Maximum output frequency: 2.7GHz
- Additive phase jitter, RMS: 0.114ps (typical) @ 156.25MHz
- Output skew: 50ps (maximum)
- LVPECL and HSTL mode operating voltage supply range:  
 $V_{CC} = 2.5V \pm 5\%$  or  $3.3V \pm 5\%$ ,  $V_{EE} = 0V$
- ECL mode operating voltage supply range:  
 $V_{EE} = -3.3V \pm 5\%$  or  $-2.5V \pm 5\%$ ,  $V_{CC} = 0V$
- $-40^{\circ}C$  to  $85^{\circ}C$  ambient operating temperature
- Lead-free (RoHS 6) packaging

### Block Diagram



### Pin Assignment

V <sub>CC</sub>	1	20	V <sub>CC</sub>
nc	2	19	Q0
V <sub>CC</sub>	3	18	nQ0
CLK_SEL	4	17	Q1
CLKA	5	16	nQ1
nCLKA	6	15	Q2
CLKB	7	14	nQ2
nCLKB	8	13	Q3
V <sub>EE</sub>	9	12	nQ3
V <sub>CC</sub>	10	11	V <sub>CC</sub>

#### IDT8T33FS314I

##### 20-Lead 209-MIL SSOP

5.3mm x 7.2mm x 1.75mm body package

##### PY Package

Top View

##### 20-Lead TSSOP

4.4mm x 6.5mm x 0.925mm body package

##### PG Package

Top View

## Pin Descriptions and Characteristics

**Table 1. Pin Descriptions**

Number	Name	Type		Description
1, 3, 10 11, 20	V <sub>CC</sub>	Power		Positive supply pins.
2	nc	Unused		No connect.
4	CLK_SEL	Input	Pulldown	Clock select input. When HIGH, selects CLKB, nCLKB inputs. When LOW, selects CLKA, nCLKA inputs.
5	CLKA	Input	Pulldown	Default non-inverting differential clock input. LVPECL/ECL interface levels.
6	nCLKA	Input	Pullup/ Pulldown	Default inverting differential clock input. LVPECL/ECL interface levels.
7	CLKB	Input	Pulldown	Alternative non-inverting differential clock input. HSTL interface levels.
8	nCLKB	Input	Pullup/ Pulldown	Alternative inverting differential clock input. HSTL interface levels.
9	V <sub>EE</sub>	Power		Negative supply pin.
12, 13	nQ3,Q3	Output		Differential output pair. LVPECL/ECL interface levels.
14, 15	nQ2,Q2	Output		Differential output pair. LVPECL/ECL interface levels.
16, 17	nQ1,Q1	Output		Differential output pair. LVPECL/ECL interface levels.
18, 19	nQ0,Q0	Output		Differential output pair. LVPECL/ECL interface levels.

NOTE: *Pullup* and *Pulldown* refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

**Table 2. Pin Characteristics**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C <sub>IN</sub>	Input Capacitance	CLK_SEL		2		pF
R <sub>PULLUP</sub>	Input Pullup Resistor			75		kΩ
R <sub>PULLDOWN</sub>	Input Pulldown Resistor			75		kΩ

## Absolute Maximum Ratings

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of the product at these conditions or any conditions beyond those listed in the DC Characteristics or AC Characteristics is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, $V_{CC}$	3.9V (LVPECL mode, $V_{EE} = 0V$ )
Negative Supply Voltage, $V_{EE}$	-3.9V (ECL mode, $V_{CC} = 0V$ )
Inputs, $V_I$ (LVPECL mode)	-0.3V to $V_{CC} + 0.3V$
Inputs, $V_I$ (ECL mode)	0.3V to $V_{EE} - 0.3V$
Outputs, $I_O$ Continuous Current	50mA
Package Thermal Impedance, $\theta_{JA}$ 20-Lead SSOP 20-Lead TSSOP	100.4°C/W (0 mps) 115.0°C/W (0 mps)
Storage Temperature, $T_{STG}$	-65°C to 150°C

## DC Electrical Characteristics

**Table 3A. LVPECL DC Characteristics,  $V_{CC} = 2.5V \pm 5\%$  or  $3.3V \pm 5\%$ ,  $V_{EE} = 0V$ ,  $T_A = -40^\circ C$  to  $85^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Control Input CLK_SEL						
V <sub>IL</sub>	Input Low Voltage		V <sub>CC</sub> - 1.810		V <sub>CC</sub> - 1.475	V
V <sub>IH</sub>	Input High Voltage		V <sub>CC</sub> - 1.165		V <sub>CC</sub> - 0.880	V
I <sub>IN</sub>	Input Current	V <sub>IN</sub> = V <sub>IL</sub> or V <sub>IN</sub> = V <sub>IH</sub>			100	μA
Clock Input Pair CLKA, nCLKA (LVPECL Differential Signals)						
V <sub>PP</sub>	Peak-to-Peak Input Voltage; NOTE 1		0.1		1.3	V
V <sub>CMR</sub>	Common Mode Input Voltage; NOTE 2		1.0		V <sub>CC</sub> - 0.3	V
I <sub>IN</sub>	Input Current	V <sub>IN</sub> = V <sub>IL</sub> or V <sub>IN</sub> = V <sub>IH</sub>			100	μA
Clock Input Pair CLKB, nCLKB (HSTL Differential Signals)						
V <sub>DIF</sub>	Differential Input Voltage; NOTE 3	V <sub>CC</sub> = 3.3V	0.4			V
		V <sub>CC</sub> = 2.5V	0.4			V
V <sub>X</sub>	Differential Crosspoint Voltage; NOTE 4		0	0.68 - 0.9	V <sub>CC</sub> - 1.0	V
I <sub>IN</sub>	Input Current	V <sub>IN</sub> = V <sub>x</sub> ±0.2V			200	μA
LVPECL Clock Outputs (Q[0:3], nQ[0:3])						
V <sub>OH</sub>	Output High Voltage		V <sub>CC</sub> - 1.2	V <sub>CC</sub> - 0.808	V <sub>CC</sub> - 0.7	V
V <sub>OL</sub>	Output Low Voltage	V <sub>CC</sub> = 3.3V ±5%	V <sub>CC</sub> - 1.9	V <sub>CC</sub> - 1.689	V <sub>CC</sub> - 1.5	V
		V <sub>CC</sub> = 2.5V ±5%	V <sub>CC</sub> - 1.9	V <sub>CC</sub> - 1.662	V <sub>CC</sub> - 1.3	V
Supply Current						
I <sub>EE</sub>	Maximum Quiescent Supply Current without Output Termination Current			49	60	mA

NOTE 1:  $V_{PP}$  is the minimum differential input voltage swing required to maintain device functionality.

NOTE 2:  $V_{CMR}$  is the crosspoint of the differential input signal. Functional operation is obtained when the crosspoint is within the  $V_{CMR}$  range and the input swing lies within the  $V_{PP}$  specification.

NOTE 3:  $V_{DIF}$  is the minimum differential HSTL input voltage swing required for device functionality.

NOTE 4:  $V_X$  is the crosspoint of the differential HSTL input signal. Functional operation is obtained when the crosspoint is within the  $V_X$  range and the input swing lies within the  $V_{PP}$  specification.

**Table 3B. ECL DC Characteristics,  $V_{CC} = 0V$ ,  $V_{EE} = -3.3V \pm 5\%$  or  $-2.5V \pm 5\%$ ,  $T_A = -40^\circ C$  to  $85^\circ C$** 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Control Input CLK_SEL						
V <sub>IL</sub>	Input Low Voltage		-1.810		-1.475	V
V <sub>IH</sub>	Input High Voltage		-1.165		-0.880	V
I <sub>IN</sub>	Input Current	V <sub>IN</sub> = V <sub>IL</sub> or V <sub>IN</sub> = V <sub>IH</sub>			100	μA
Clock Input Pair CLKA, nCLKA (LVPECL Differential Signals)						
V <sub>PP</sub>	Peak-to-Peak Input Voltage; NOTE 1		0.1		1.3	V
V <sub>CMR</sub>	Common Mode Input Voltage; NOTE 2		V <sub>EE</sub> + 1.0		-0.3	V
I <sub>IN</sub>	Input Current	V <sub>IN</sub> = V <sub>IL</sub> or V <sub>IN</sub> = V <sub>IH</sub>			100	μA
LVPECL Clock Outputs (Q[0:3], nQ[0:3])						
V <sub>OH</sub>	Output High Voltage		-1.2	-0.808	-0.7	V
V <sub>OL</sub>	Output Low Voltage	V <sub>EE</sub> = -3.3V ±5%	-1.9	-1.689	-1.5	V
		V <sub>EE</sub> = -2.5V ±5%	-1.9	-1.662	-1.3	V
Supply Current						
I <sub>EE</sub>	Maximum Quiescent Supply Current without Output Termination Current			49	60	mA

NOTE 1:  $V_{PP}$  is the minimum differential input voltage swing required to maintain device functionality.

NOTE 2:  $V_{CMR}$  is the crosspoint of the differential input signal. Functional operation is obtained when the crosspoint is within the  $V_{CMR}$  range and the input swing lies within the  $V_{PP}$  specification.

## AC Electrical Characteristics

**Table 4. AC Characteristics, (LVPECL/HSTL),  $V_{CC} = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$ ,  $V_{EE} = 0V$ , or (ECL)  $V_{EE} = -3.3V \pm 5\%$  or  $-2.5V \pm 5\%$ ,  $V_{CC} = 0V$ ,  $T_A = -40^\circ C$  to  $85^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{PP}$	Differential Input Voltage; NOTE 1		0.15		1.3	V
$V_{CMR}$	Differential Input Crosspoint Voltage; NOTE 2		$V_{EE} + 1.0$		$V_{CC} - 0.3$	V
$f_{CLK}$	Input Frequency; NOTE 3				2.7	GHz
$t_{PD}$	Propagation Delay, CLKA or CLKB to Output Pair; NOTE 4		230		650	ps
$V_{DIF}$	HSTL Differential Input Voltage; NOTE 5		0.4		1.0	V
$V_X$	HSTL Input Differential Crosspoint Voltage; NOTE 6		$V_{EE} + 0.01$		$V_{CC} - 1.0$	V
$V_{O(pp)}$	Differential Output Voltage (peak-to-peak)	$f_{OUT} < 300MHz$	0.45	0.88	0.95	V
		$f_{OUT} < 1.5GHz$	0.3	0.74	0.95	V
$tsk(o)$	Output Skew				50	ps
$t_{jit}$	Buffer Additive Phase Jitter, RMS; refer to Additive Phase Jitter Section	156.25MHz, @ 3.3V, 1.875MHz – 20MHz		0.114		ps
		312.5MHz @ 3.3V, 1.875MHz – 20MHz		0.052		ps
$tsk(p)$	Output Pulse Skew; NOTE 7	660MHz			75	ps
$t_R / t_F$	Output Rise/Fall Time	20% to 80%	0.05		0.3	ns

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1:  $V_{PP}$  is the minimum differential ECL/LVPECL input voltage swing required to maintain AC characteristics including  $t_{PD}$  and device-to-device skew.

NOTE 2:  $V_{CMR}$  is the crosspoint of the differential ECL/LVPECL input signal. Normal AC operation is obtained when the crosspoint is within the  $V_{CMR}$  range and the input swing lies within the  $V_{PP}$  specification. Violation of  $V_{CMR}$  or  $V_{PP}$  impacts the device propagation delay, device and part-to-part skew.

NOTE 3: The IDT8T33FS314 is fully operational up to 2.7GHz and is characterized up to 1.5GHz.

NOTE 4: Propagation delay specified for output rise and fall times less than 5ns.

NOTE 5:  $V_{DIF}$  is the minimum differential HSTL input voltage swing required to maintain AC characteristics including  $t_{PD}$  and device-to-device skew.

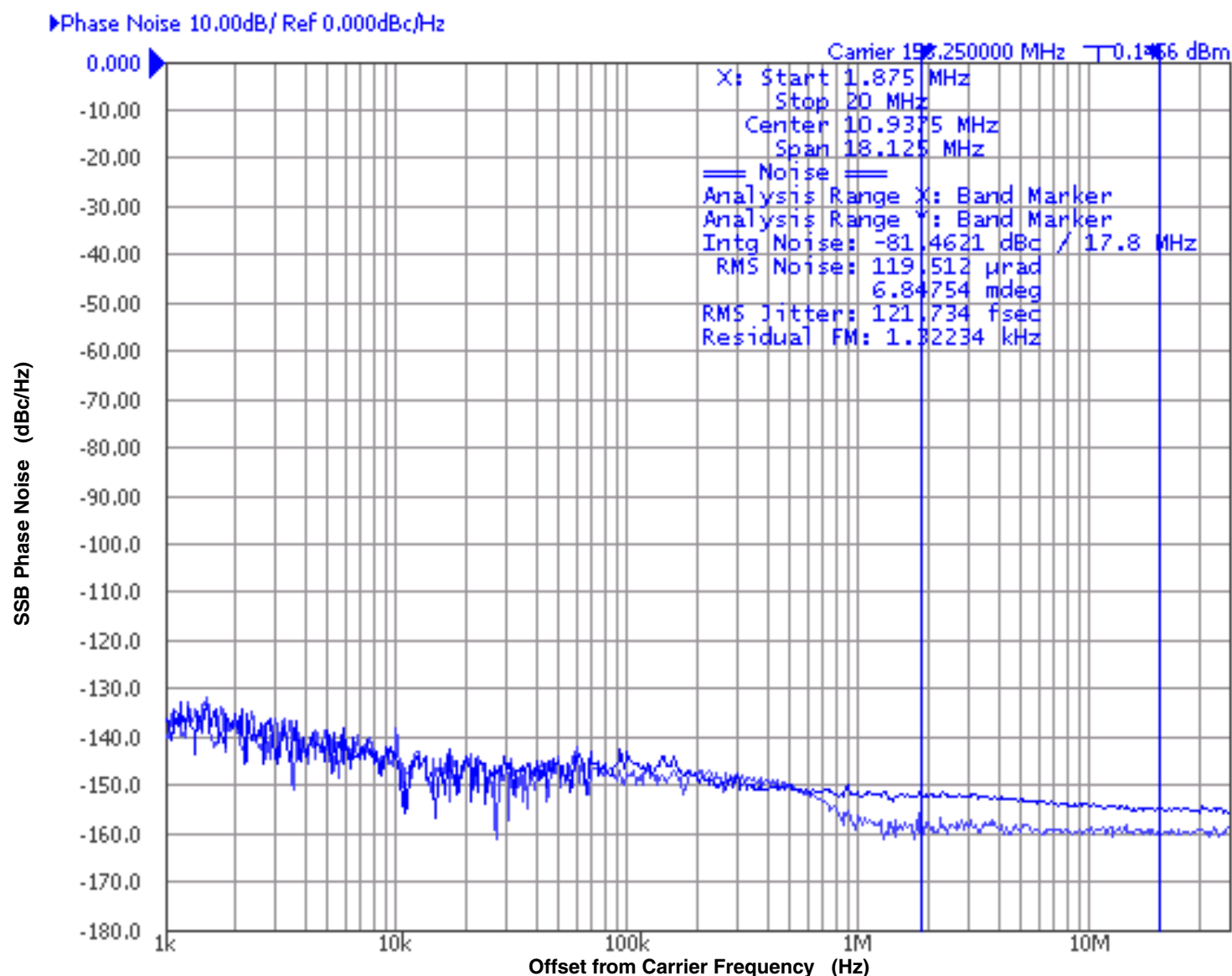
NOTE 6:  $V_X$  is the crosspoint of the differential HSTL input signal. Normal AC operation is obtained when the crosspoint is within the  $V_X$  range and the input swing lies within the  $V_{DIF}$  specification. Violation of  $V_X$  or  $V_{DIF}$  impacts the device propagation delay, device and part-to-part skew.

NOTE 7: Output pulse skew is the absolute value of the difference of the propagation delay times:  $|t_{PLH} - t_{PHL}|$ .

## Additive Phase Jitter

The spectral purity in a band at a specific offset from the fundamental compared to the power of the fundamental is called the **dBc Phase Noise**. This value is normally expressed using a Phase noise plot and is most often the specified plot in many applications. Phase noise is defined as the ratio of the noise power present in a 1Hz band at a specified offset from the fundamental frequency to the power value of the fundamental. This ratio is expressed in decibels (dBm) or a ratio

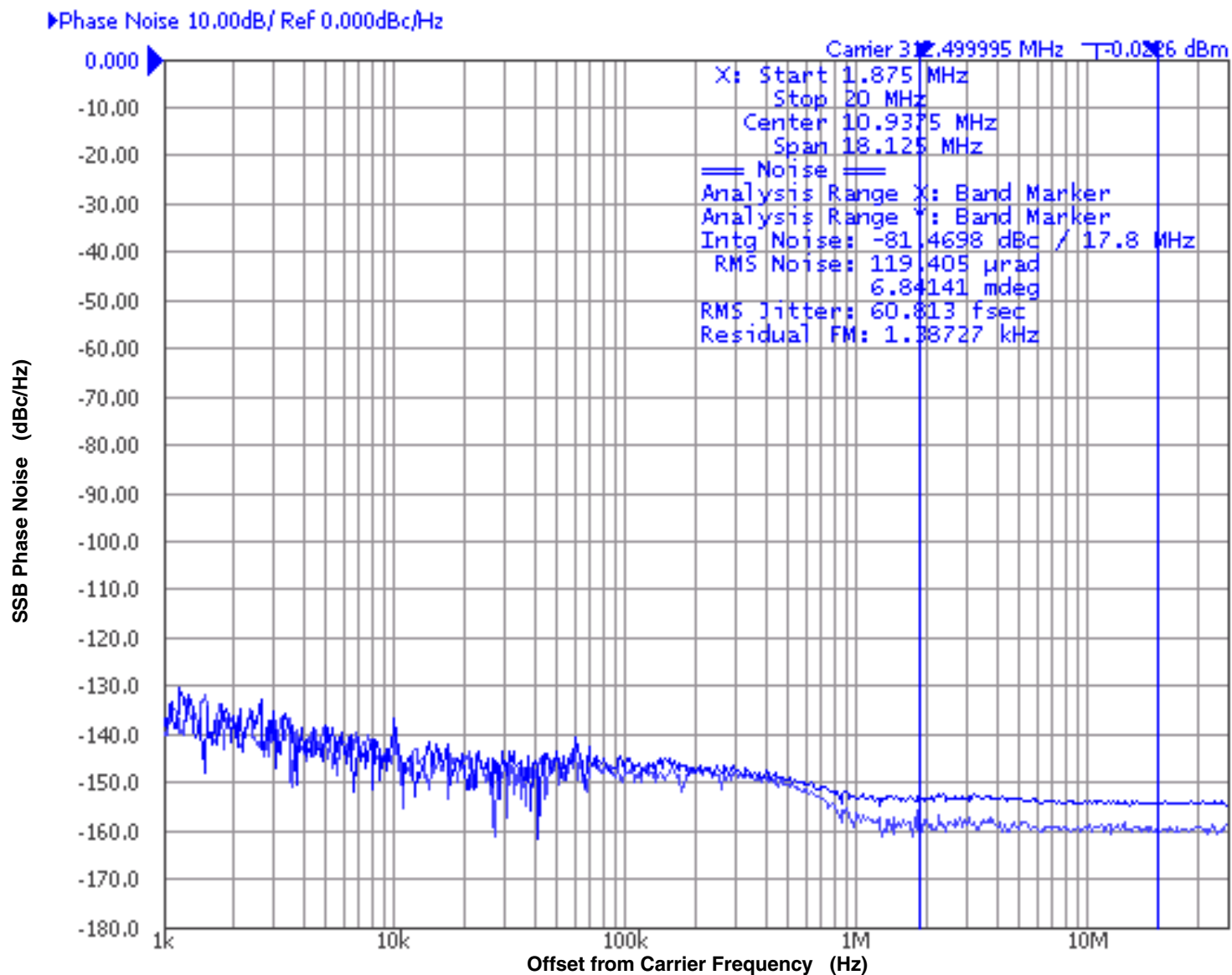
of the power in the 1Hz band to the power in the fundamental. When the required offset is specified, the phase noise is called a **dBc** value, which simply means dBm at a specified offset from the fundamental. By investigating jitter in the frequency domain, we get a better understanding of its effects on the desired application over the entire time record of the signal. It is mathematically possible to calculate an expected bit error rate given a phase noise plot.



As with most timing specifications, phase noise measurements have issues relating to the limitations of the equipment. Often the noise floor of the equipment is higher than the noise floor of the device. This is illustrated above. The device meets the noise floor of what is

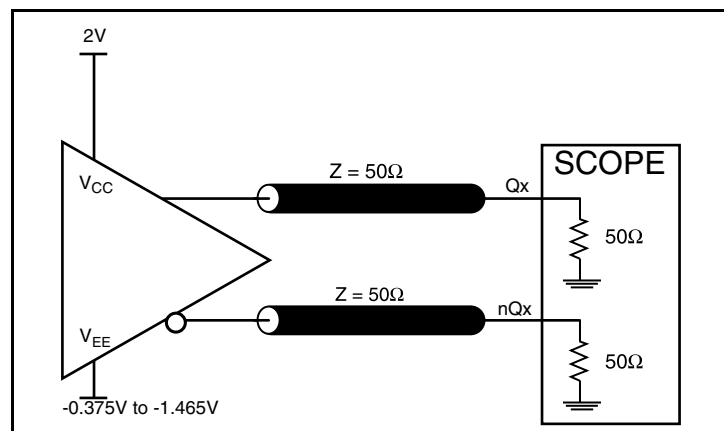
shown, but can actually be lower. The phase noise is dependent on the input source and measurement equipment.

A Rohde & Schartz SMA100 was used as input source.

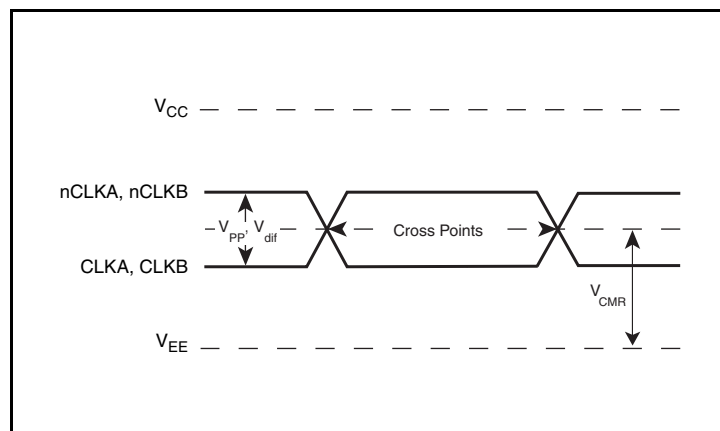


A Rohde & Schartz SMA100 was used as input source.

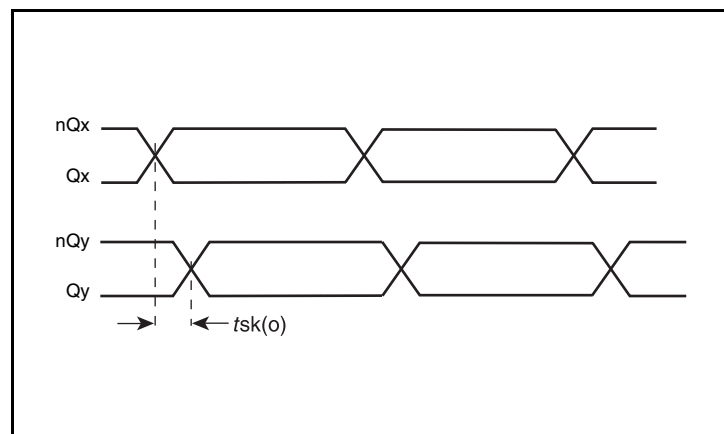
# Parameter Measurement Information



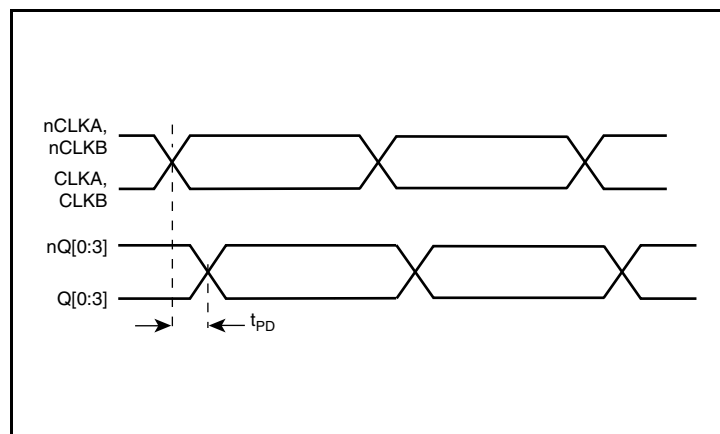
3.3V, 2.5V LVPECL Output Load Test Circuit



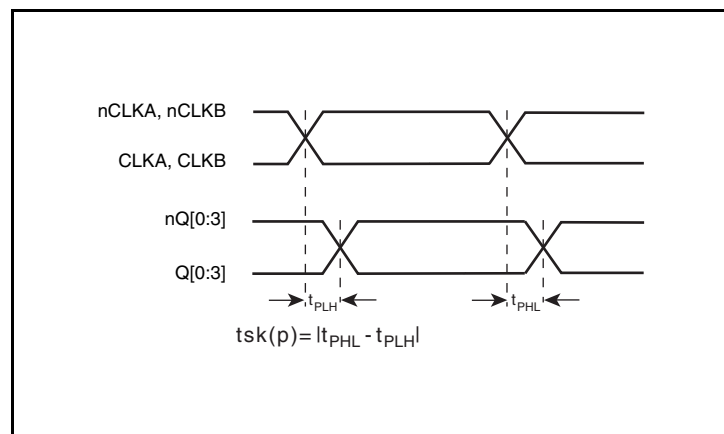
Differential Input Level



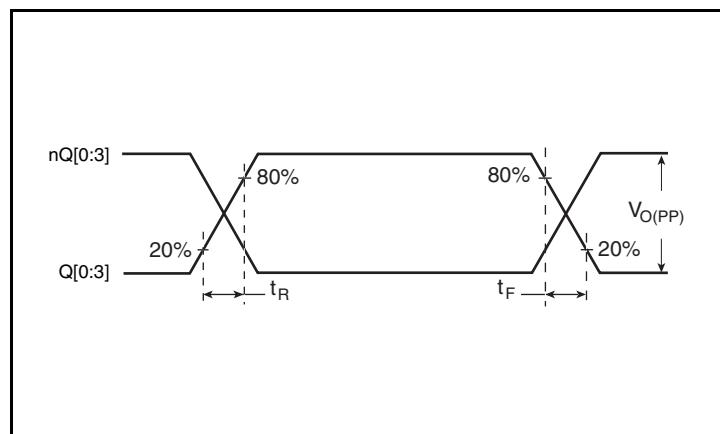
Output Skew



Propagation Delay



Output Pulse Skew



Output Rise/Fall Time



## Applications Information

### Recommendations for Unused Input and Output Pins

#### Inputs:

##### CLK/nCLK Inputs

For applications not requiring the use of a differential input, both the CLK and nCLK pins can be left floating. Though not required, but for additional protection, a 1k $\Omega$  resistor can be tied from CLK to ground.

#### Outputs:

##### LVPECL Outputs

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

### Wiring the Differential Input to Accept Single-Ended Levels

Figure 1 shows how a differential input can be wired to accept single ended levels. The reference voltage  $V_1 = V_{CC}/2$  is generated by the bias resistors R1 and R2. The bypass capacitor (C1) is used to help filter noise on the DC bias. This bias circuit should be located as close to the input pin as possible. The ratio of R1 and R2 might need to be adjusted to position the  $V_1$  in the center of the input voltage swing. For example, if the input clock swing is 2.5V and  $V_{CC} = 3.3V$ , R1 and R2 value should be adjusted to set  $V_1$  at 1.25V. Similarly, if the input clock swing is 1.8V and  $V_{CC} = 3.3V$ , R1 and R2 value should be adjusted to set  $V_1$  at 0.9V. It is recommended to always use R1 and R2 to provide a known  $V_1$  voltage. The values below are for when both the single ended swing and  $V_{CC}$  are at the same voltage. This configuration requires that the sum of the output impedance of the driver ( $R_o$ ) and the series resistance ( $R_s$ ) equals the transmission line impedance. In addition, matched termination at the input will attenuate the signal in half. This can be done in one of two ways.

First, R3 and R4 in parallel should equal the transmission line impedance. For most 50 $\Omega$  applications, R3 and R4 can be 100 $\Omega$ . The values of the resistors can be increased to reduce the loading for slower and weaker LVCMOS driver. When using single-ended signaling, the noise rejection benefits of differential signaling are reduced. Even though the differential input can handle full rail LVCMOS signaling, it is recommended that the amplitude be reduced. The datasheet specifies a lower differential amplitude, however this only applies to differential signals. For single-ended applications, the swing can be larger, however  $V_{IL}$  cannot be less than -0.3V and  $V_{IH}$  cannot be more than  $V_{CC} + 0.3V$ . Though some of the recommended components might not be used, the pads should be placed in the layout. They can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a differential signal.

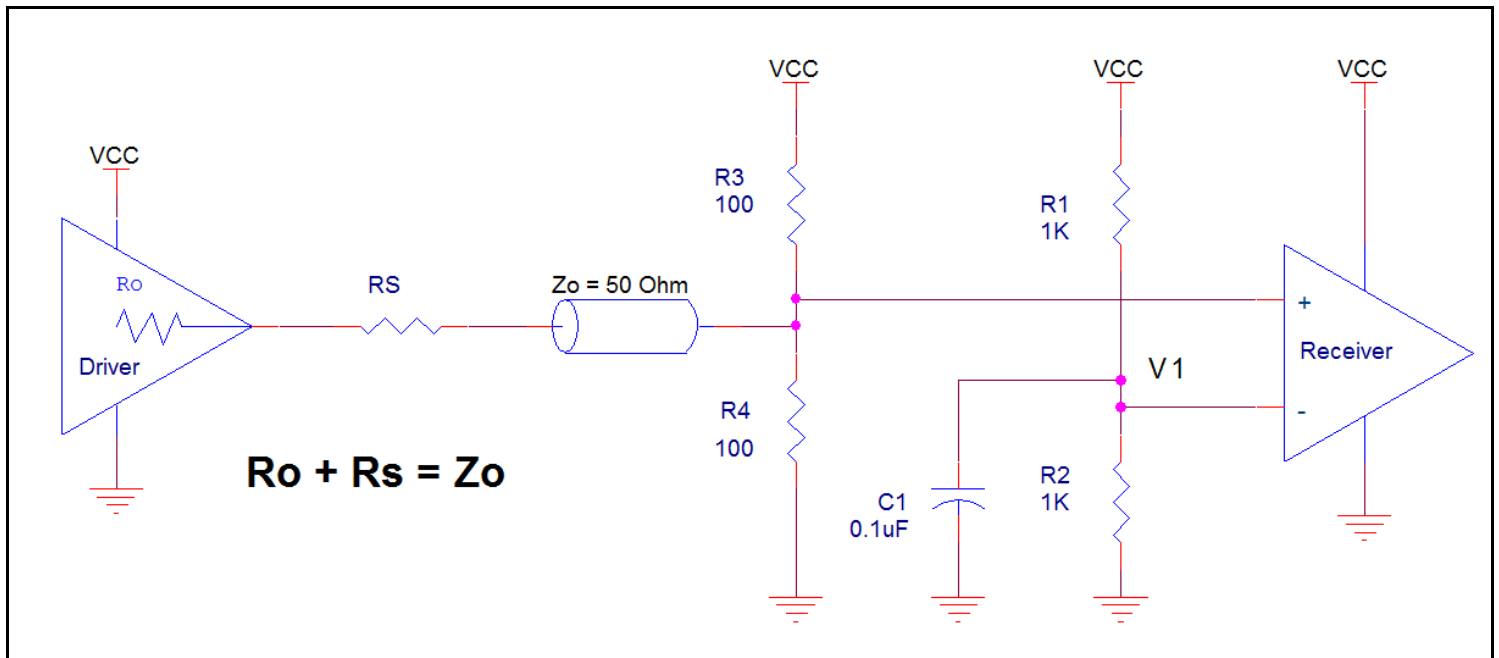
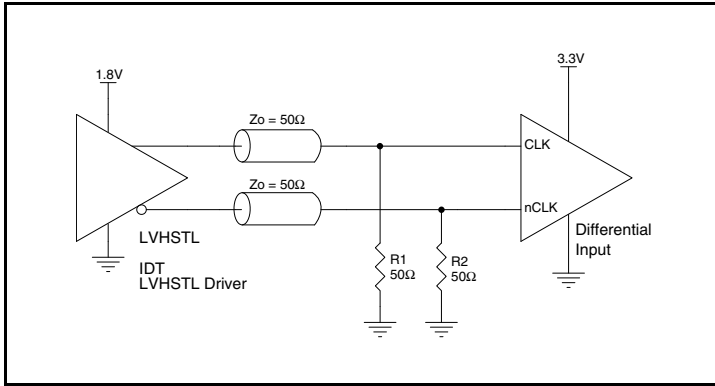


Figure 1. Recommended Schematic for Wiring a Differential Input to Accept Single-ended Levels

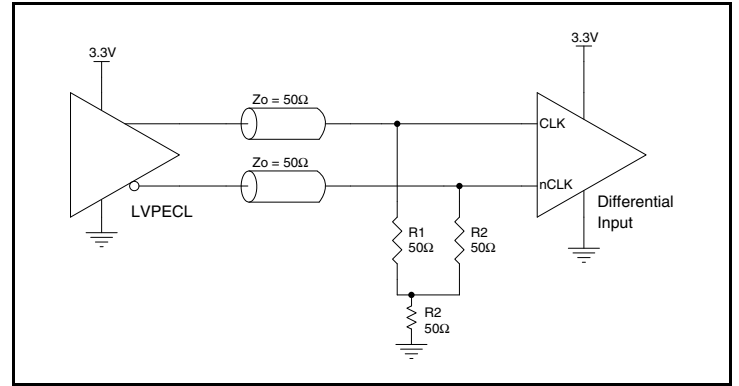
### 3.3V Differential Clock Input Interface

CLKx/nCLKx accepts LVDS, LVPECL, LVHSTL, HCSL and other differential signals. Both  $V_{O(PP)}$  and  $V_{OH}$  must meet the  $V_{PP}$  and  $V_{CMR}$  input requirements. Figure 2A to Figure 2E show interface examples for the CLKx/nCLKx input driven by the most common driver types. The input interfaces suggested here are examples only.

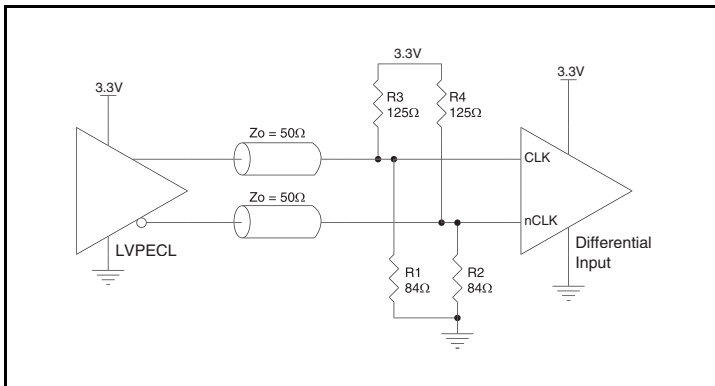
Please consult with the vendor of the driver component to confirm the driver termination requirements. For example, in Figure 2A, the input termination applies for IDT open emitter LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.



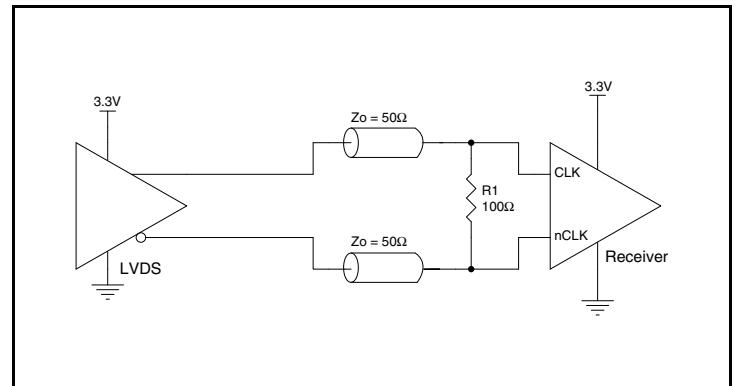
**Figure 2A. CLKx/nCLKx Input Driven by an IDT Open Emitter LVHSTL Driver**



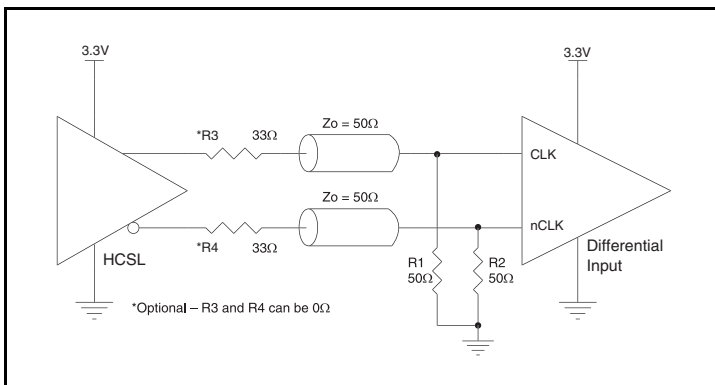
**Figure 2B. CLKx/nCLKx Input Driven by a 3.3V LVPECL Driver**



**Figure 2C. CLKx/nCLKx Input Driven by a 3.3V LVPECL Driver**



**Figure 2D. CLKx/nCLKx Input Driven by a 3.3V LVDS Driver**

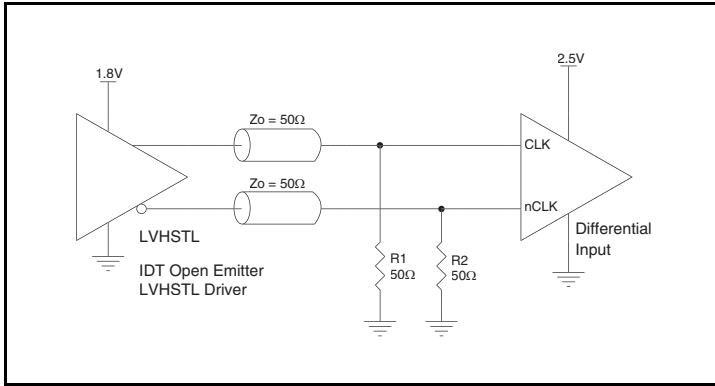


**Figure 2E. CLKx/nCLKx Input Driven by a 3.3V HCSL Driver**

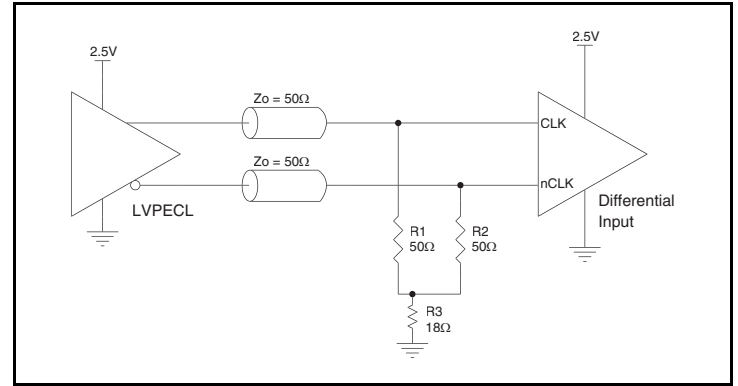
## 2.5V Differential Clock Input Interface

CLKx/nCLKx accepts LVDS, LVPECL, LVHSTL, HCSL and other differential signals. Both  $V_{O(PP)}$  and  $V_{OH}$  must meet the  $V_{PP}$  and  $V_{CMR}$  input requirements. *Figure 3A to Figure 3E* show interface examples for the CLKx/nCLKx input driven by the most common driver types. The input interfaces suggested here are examples only.

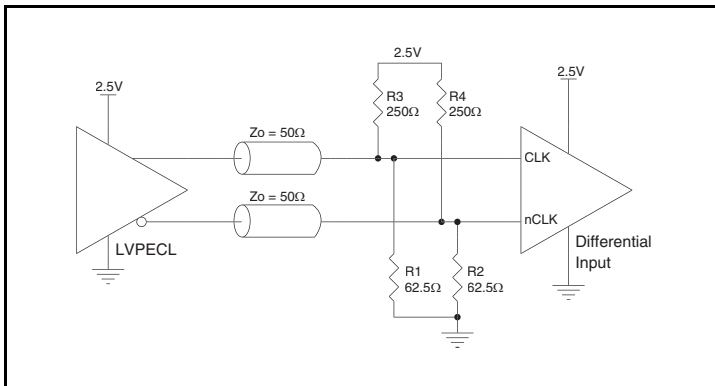
Please consult with the vendor of the driver component to confirm the driver termination requirements. For example, in *Figure 3A*, the input termination applies for IDT open emitter LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.



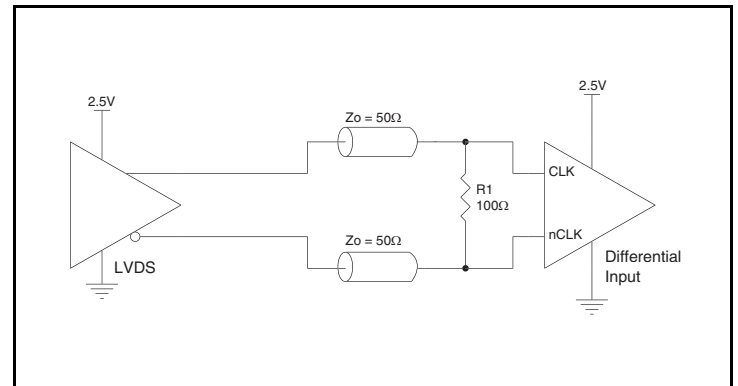
**Figure 3A.** CLKx/nCLKx Input Driven by an IDT Open Emitter LVHSTL Driver



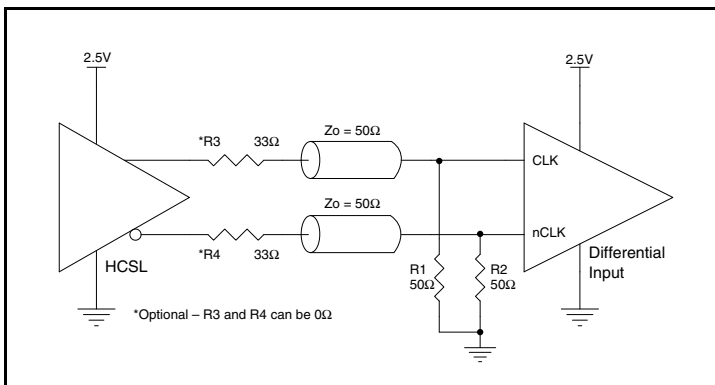
**Figure 3B.** CLKx/nCLKx Input Driven by a 2.5V LVPECL Driver



**Figure 3C.** CLKx/nCLKx Input Driven by a 2.5V LVPECL Driver



**Figure 3D.** CLKx/nCLKx Input Driven by a 2.5V LVDS Driver



**Figure 3E.** CLKx/nCLKx Input Driven by a 2.5V HCSL Driver

## Termination for 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are a low impedance follower output that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion.

*Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

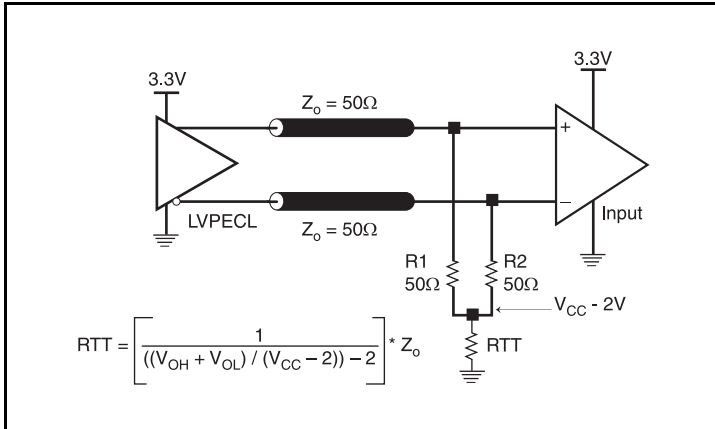


Figure 4A. 3.3V LVPECL Output Termination

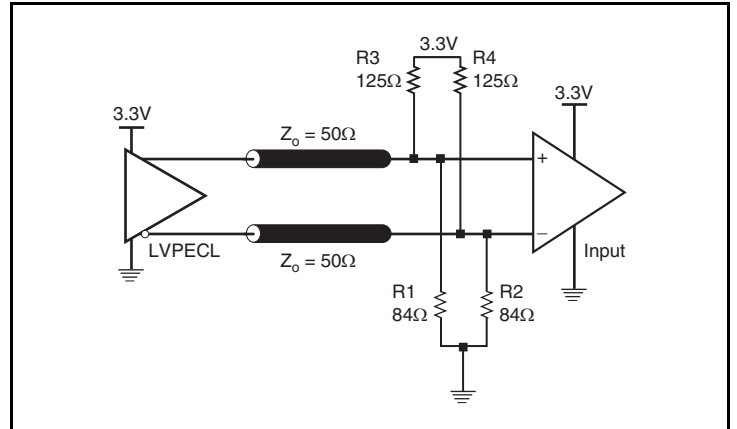


Figure 4B. 3.3V LVPECL Output Termination

## Termination for 2.5V LVPECL Outputs

Figure 5A and Figure 5B show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating  $50\Omega$  to  $V_{CC} - 2V$ . For  $V_{CC} = 2.5V$ , the  $V_{CC} - 2V$  is very close to ground

level. The  $R3$  in Figure 5B can be eliminated and the termination is shown in Figure 5C.

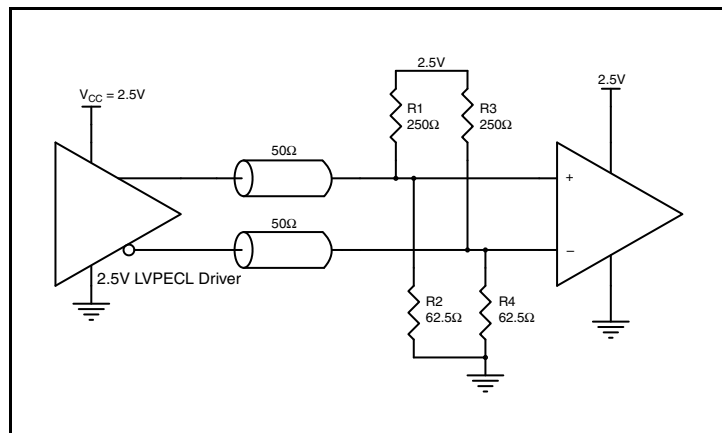


Figure 5A. 2.5V LVPECL Driver Termination Example

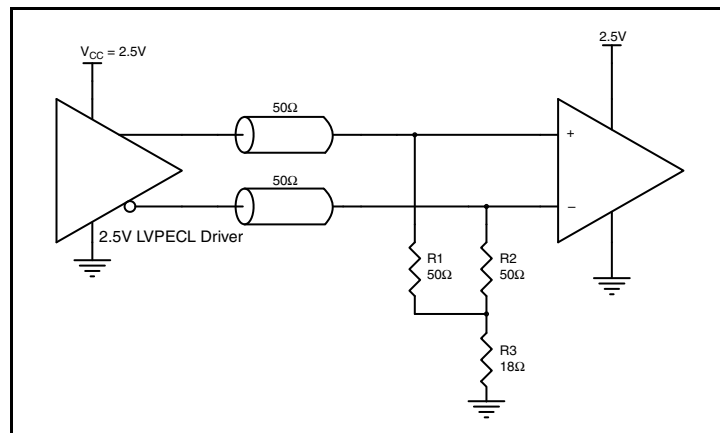


Figure 5B. 2.5V LVPECL Driver Termination Example

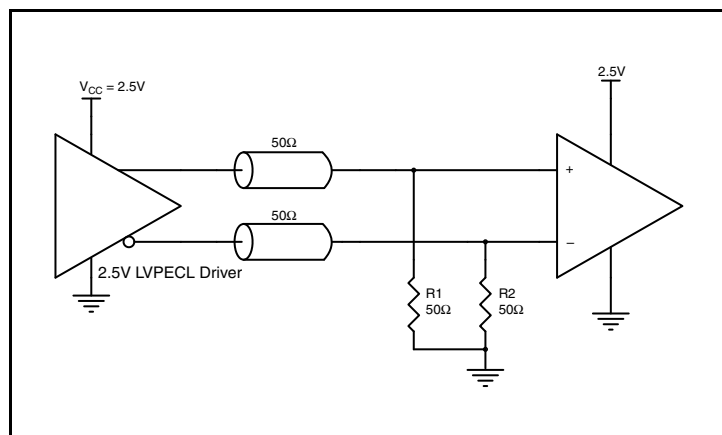


Figure 5C. 2.5V LVPECL Driver Termination Example

## Power Considerations

This section provides information on power dissipation and junction temperature for the IDT8T33FS314I. Equations and example calculations are also provided.

### 1. Power Dissipation.

The total power dissipation for the IDT8T33FS314I is the sum of the core power plus the power dissipated due to the load. The following is the power dissipation for  $V_{CC} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)<sub>MAX</sub> =  $V_{CC\_MAX} * I_{EE\_MAX} = 3.465V * 60mA = \mathbf{207.9mW}$
  - Power (outputs)<sub>MAX</sub> = **33.2mW/Loaded Output pair**  
If all outputs are loaded, the total power is  $4 * 33.2mW = \mathbf{132.8mW}$
- Total Power**<sub>MAX</sub> (3.465V, with all outputs switching) =  $207.9mW + 132.8mW = \mathbf{340.7mW}$

### 2. Junction Temperature.

Junction temperature,  $T_j$ , is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature,  $T_j$ , to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for  $T_j$  is as follows:  $T_j = \theta_{JA} * Pd\_total + T_A$

$T_j$  = Junction Temperature

$\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

$Pd\_total$  = Total Device Power Dissipation (example calculation is in section 1 above)

$T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is 115°C/W per Table 5A below.

Therefore,  $T_j$  for an ambient temperature of 85°C with all outputs switching is:

$$85^\circ C + 0.341W * 115^\circ C/W = 124.2^\circ C. \text{ This is within the limit of } 125^\circ C.$$

This calculation is only an example.  $T_j$  will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

**Table 5A. Thermal Resistance  $\theta_{JA}$  for 20-Lead TSSOP, Forced Convection**

$\theta_{JA}$ by Velocity			
Meters per Second	0	1	2
Multi-Layer PCB, JEDEC Standard Test Boards	115.0°C/W	109.1°C/W	103.9°C/W

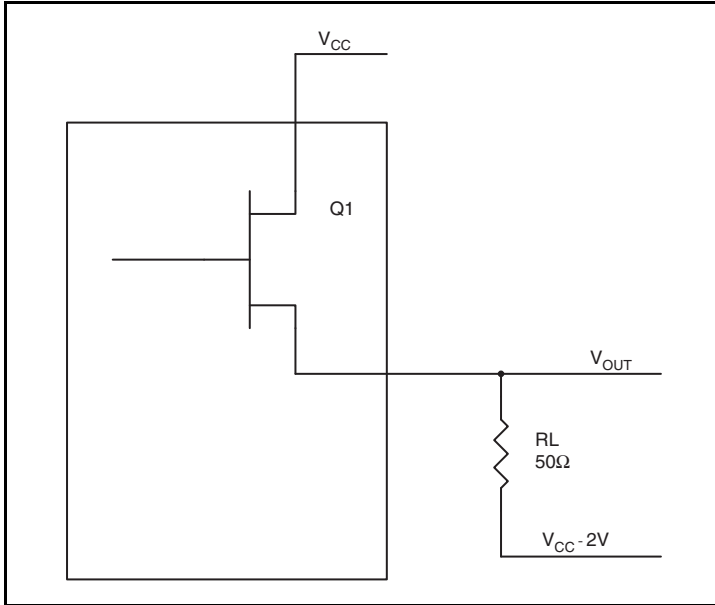
**Table 5B. Thermal Resistance  $\theta_{JA}$  for 20-Lead SSOP, Forced Convection**

$\theta_{JA}$ by Velocity			
Meters per Second	0	1	2
Multi-Layer PCB, JEDEC Standard Test Boards	100.4°C/W	92.7°C/W	87.3°C/W

### 3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pairs.

LVPECL output driver circuit and termination are shown in *Figure 6*.



**Figure 6. LVPECL Driver Circuit and Termination**

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of  $V_{CC} - 2V$ .

- For logic high,  $V_{OUT} = V_{OH\_MAX} = V_{CC\_MAX} - 0.7V$   
( $V_{CC\_MAX} - V_{OH\_MAX}$ ) = **0.7V**
- For logic low,  $V_{OUT} = V_{OL\_MAX} = V_{CC\_MAX} - 1.5V$   
( $V_{CC\_MAX} - V_{OL\_MAX}$ ) = **1.5V**

$Pd\_H$  is power dissipation when the output drives high.

$Pd\_L$  is the power dissipation when the output drives low.

$$Pd\_H = [(V_{OH\_MAX} - (V_{CC\_MAX} - 2V))/R_L] * (V_{CC\_MAX} - V_{OH\_MAX}) = [(2V - (V_{CC\_MAX} - V_{OH\_MAX}))/R_L] * (V_{CC\_MAX} - V_{OH\_MAX}) = [(2V - 0.7V)/50\Omega] * 0.7V = \mathbf{18.2mW}$$

$$Pd\_L = [(V_{OL\_MAX} - (V_{CC\_MAX} - 2V))/R_L] * (V_{CC\_MAX} - V_{OL\_MAX}) = [(2V - (V_{CC\_MAX} - V_{OL\_MAX}))/R_L] * (V_{CC\_MAX} - V_{OL\_MAX}) = [(2V - 1.5V)/50\Omega] * 1.5V = \mathbf{15mW}$$

$$\text{Total Power Dissipation per output pair} = Pd\_H + Pd\_L = \mathbf{33.2mW}$$

## Reliability Information

**Table 6A.  $\theta_{JA}$  vs. Air Flow Table for a 20-Lead TSSOP**

$\theta_{JA}$ by Velocity			
Meters per Second	0	1	2
Multi-Layer PCB, JEDEC Standard Test Boards	115.0°C/W	109.1°C/W	103.9°C/W

**Table 6B.  $\theta_{JA}$  vs. Air Flow Table for a 20-Lead SSOP**

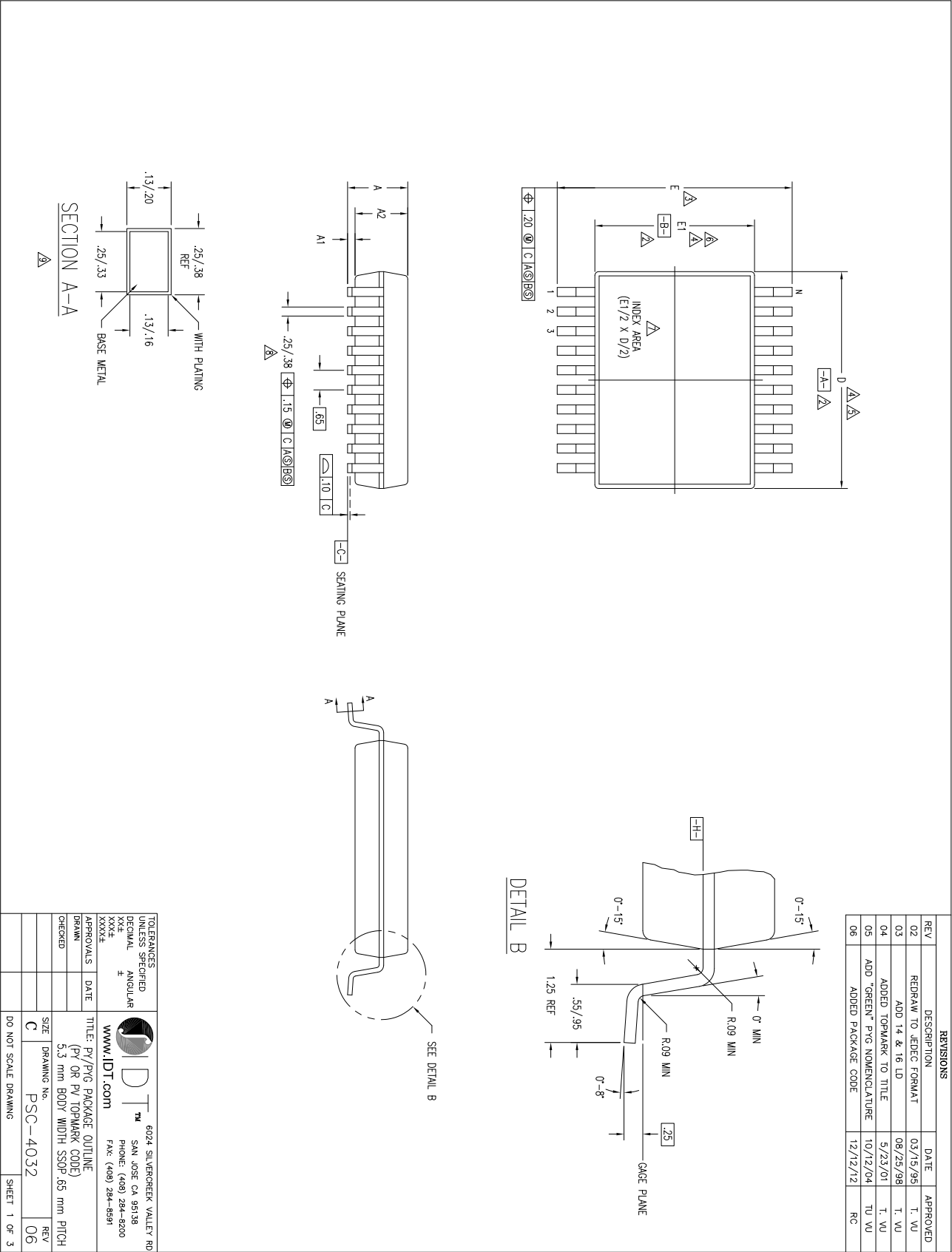
$\theta_{JA}$ by Velocity			
Meters per Second	0	1	2
Multi-Layer PCB, JEDEC Standard Test Boards	100.4°C/W	92.7°C/W	87.3°C/W

## Transistor Count

The transistor count for IDT8T33FS314I is: 882



20-Lead SSOP Package Outline and Package Dimensions








## 20-Lead SSOP Package Outline and Package Dimensions, continued

Symbol	PY/PYG14				Notes	PY/PYG16				Notes	PY/PYG20				Notes	PY/PYG24				Notes	PY/PYG28				Notes
	JEDEC VARIATION					JEDEC VARIATION					JEDEC VARIATION					JEDEC VARIATION					JEDEC VARIATION				
	MIN	NOM	MAX			MIN	NOM	MAX			MIN	NOM	MAX			MIN	NOM	MAX			MIN	NOM	MAX		
AMB	1.73	1.86	1.99			1.73	1.86	1.99			1.73	1.86	1.99			1.73	1.86	1.99			1.73	1.86	1.99		
BD	.05	.13	.21			.05	.13	.21			.05	.13	.21			.05	.13	.21			.05	.13	.21		
L	1.68	1.73	1.78			1.68	1.73	1.78			1.68	1.73	1.78			1.68	1.73	1.78			1.68	1.73	1.78		
E	5.80	6.20	6.50	4.5		5.90	6.20	6.50	4.5		7.07	7.20	7.33	4.5		8.07	8.20	8.33	4.5		10.07	10.20	10.33	4.5	
D	7.65	7.80	7.90	3		7.65	7.80	7.90	3		7.65	7.80	7.90	3		7.65	7.80	7.90	3		7.65	7.80	7.90	3	
E1	5.20	5.30	5.38	4.6		5.20	5.30	5.38	4.6		5.20	5.30	5.38	4.6		5.20	5.30	5.38	4.6		5.20	5.30	5.38	4.6	
N	14					16					20					24					28				

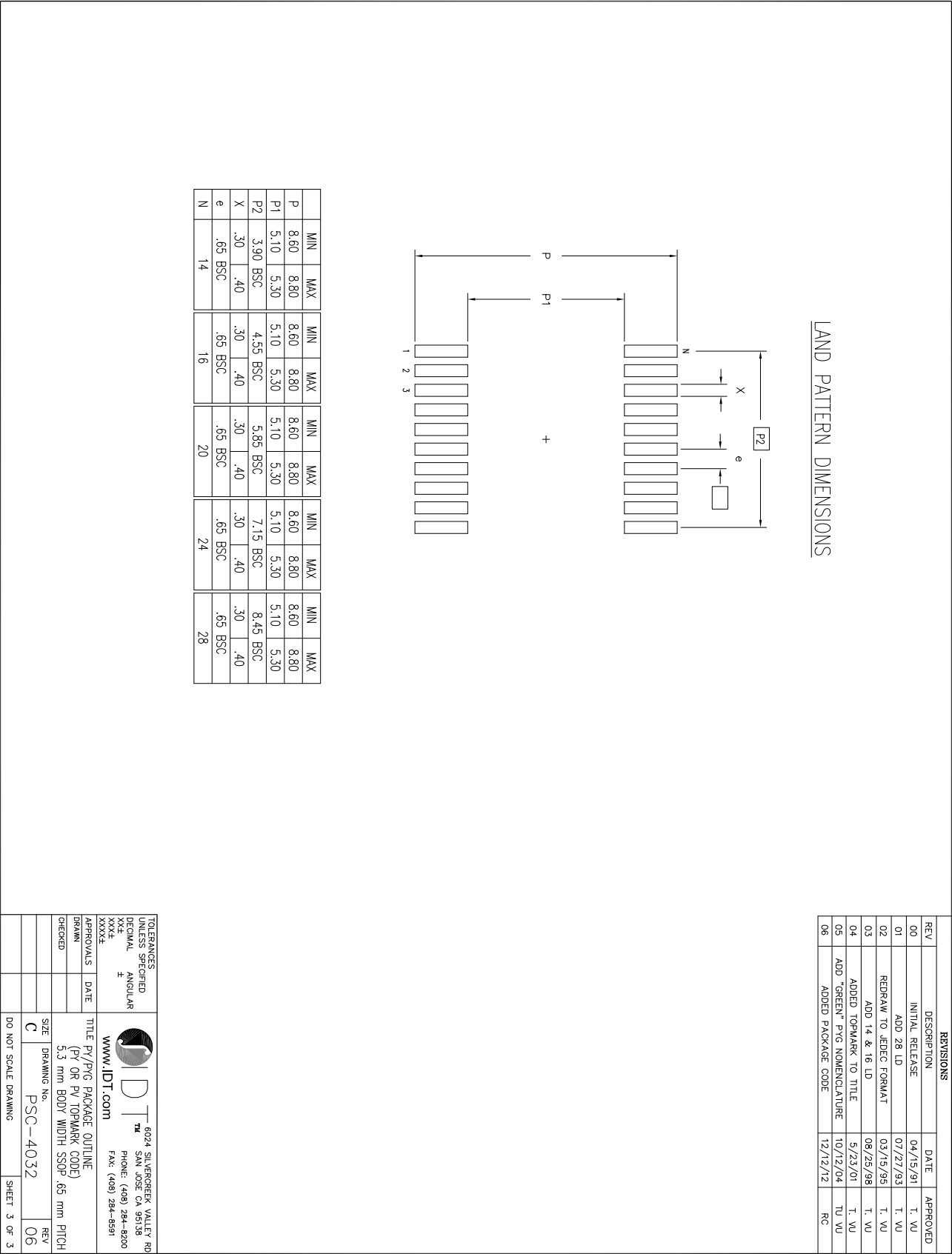
REVISIONS		
REV	DESCRIPTION	DATE
02	REDRAW TO JEDEC FORMAT	03/15/95
03	ADD 14 & 16 LD	08/25/98
04	ADDED TOPMARK TO TITLE	5/23/01
05	ADD "GREEN" PIG NOMENCLATURE	10/12/04
06	ADDED PACKAGE CODE	12/12/12

## NOTES:

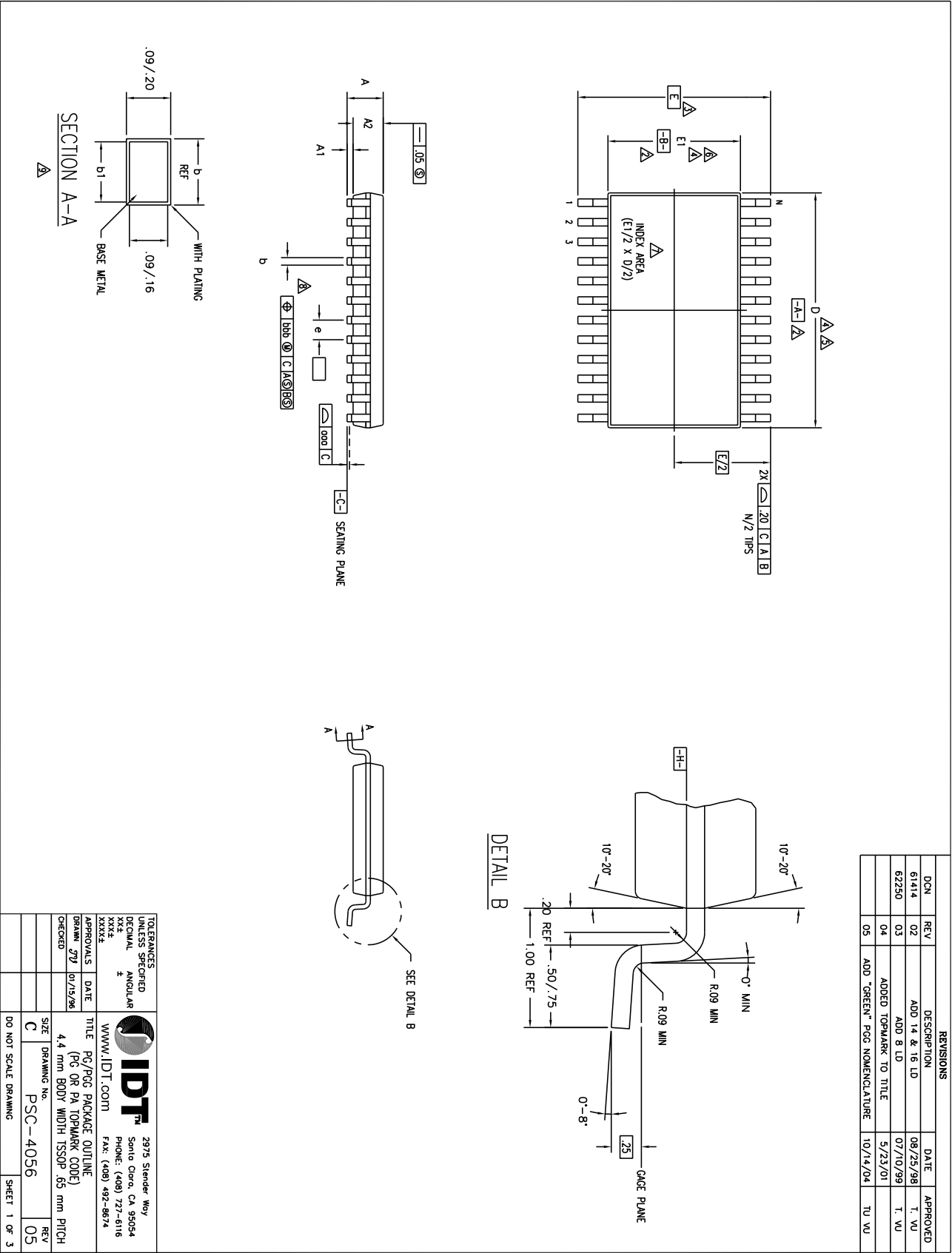
- 1 ALL DIMENSIONING AND TOLERANCING CONFORM TO ASME Y14.5M-1994
- 2 DATUMS  AND  TO BE DETERMINED AT DATUM PLANE 
- 3 DIMENSION E TO BE DETERMINED AT SEATING PLANE 
- 4 DIMENSIONS D AND E1 ARE TO BE DETERMINED AT DATUM PLANE 
- 5 DIMENSION D DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS, MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED .20 mm PER SIDE
- 6 DIMENSION E1 DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS; INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED .20 mm PER SIDE
- 7 DETAIL OF PIN 1 IDENTIFIER IS OPTIONAL BUT MUST BE LOCATED WITHIN THE ZONE INDICATED
- 8 LEAD WIDTH DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION, ALLOWABLE DAMBAR PROTRUSION IS .13 mm IN EXCESS OF THE LEAD WIDTH DIMENSION AT MAXIMUM MATERIAL CONDITION, DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT
- 9 THESE DIMENSIONS APPLY TO THE FLAT SECTION OF THE LEAD BETWEEN .10 AND .25 mm FROM THE LEAD TIP
- 10 ALL DIMENSIONS ARE IN MILLIMETERS
- 11 THIS OUTLINE CONFORMS TO JEDEC PUBLICATION 95 REGISTRATION MO-150, VARIATION AB, AC, AE, AG & AH

[illegible]

20-Lead SSOP Package Outline and Package Dimensions, continued



20-Lead TSSOP Package Outline and Package Dimensions




## 20-Lead TSSOP Package Outline and Package Dimensions, continued

	Task Reference	DATE	APPROVED
DCN	DESCRIPTION		
61414	ADD 14 & 16 LD	08/25/98	T. VU
62250	ADD 8 LD	07/03/99	T. VU
03	ADDED TOPMARK TO TITLE	5/23/01	
04			
05	ADD "GREEN" PGC NOMENCLATURE	10/14/04	TU VU

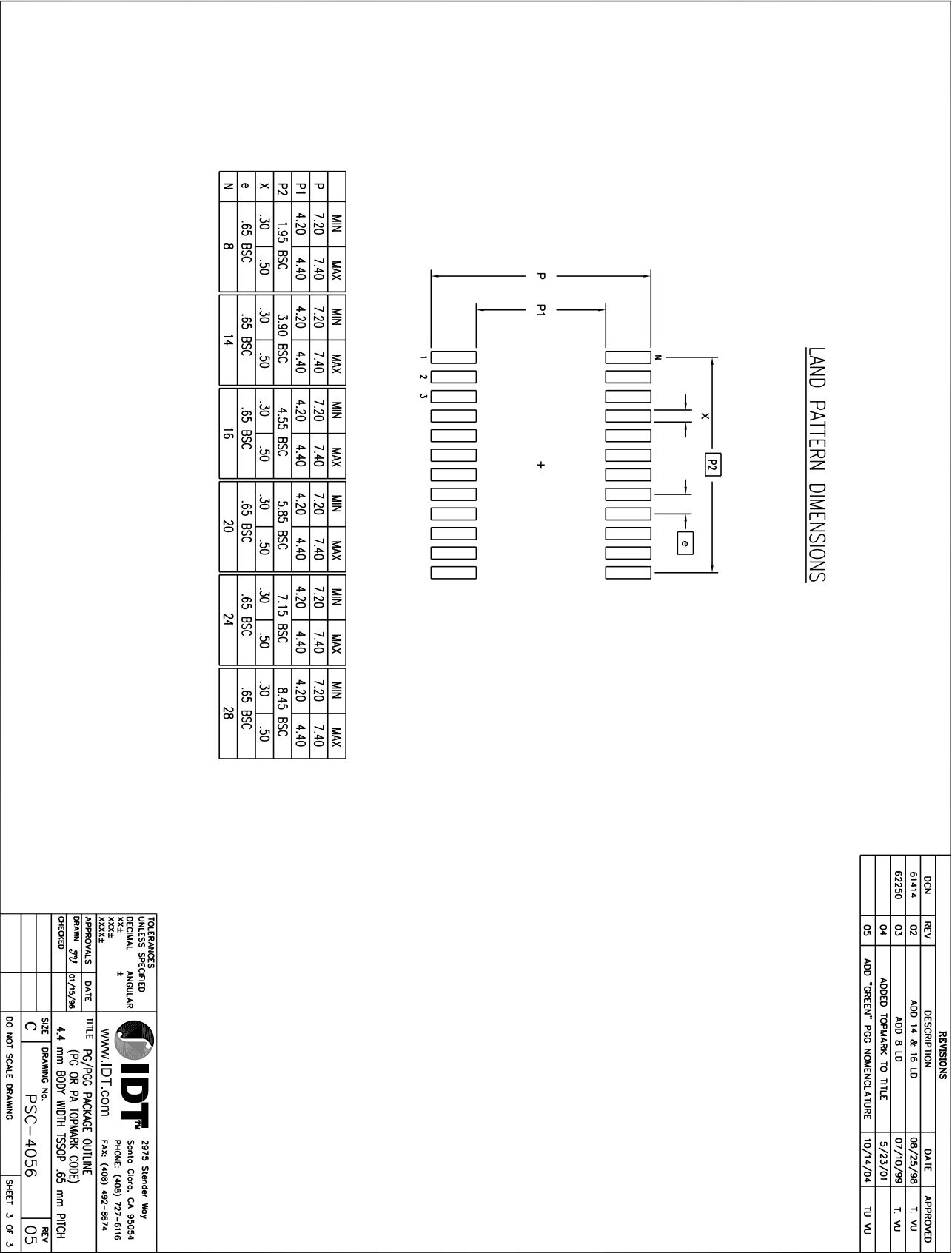
[illegible]

## NOTES:

- 1 ALL DIMENSIONING AND TOLERANCING CONFORM TO ASME Y14.5M-1994
- △ DATUMS **-A-** AND **-B-** TO BE DETERMINED AT DATUM PLANE **-H-**
- △ DIMENSION E TO BE DETERMINED AT SEATING PLANE **-C-**
- △ DIMENSIONS D AND E1 ARE TO BE DETERMINED AT DATUM PLANE **-H-**
- △ DIMENSION D DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURS, MOLD FLASH, PROTRUSIONS OR GATE BURS SHALL NOT EXCEED .15 mm PER SIDE
- △ DIMENSION E1 DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS, INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED .25 mm PER SIDE
- △ DETAIL OF PIN 1 IDENTIFIER IS OPTIONAL, BUT MUST BE LOCATED WITHIN THE ZONE INDICATED
- △ LEAD WIDTH DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION, ALLOWABLE DAMBAR PROTRUSION IS .08 mm IN EXCESS OF THE LEAD WIDTH DIMENSION AT MAXIMUM MATERIAL CONDITION. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIOUS OR THE FOOT
- △ THESE DIMENSIONS APPLY TO THE FLAT SECTION OF THE LEAD BETWEEN .10 AND .25 mm FROM THE LEAD TIP
- 10 ALL DIMENSIONS ARE IN MILLIMETERS
- 11 THIS OUTLINE CONFORMS TO JEDEC PUBLICATION 95 REGISTRATION MO-153, VARIATION AA, AB-1, AB, AC, AD & AE

TOLERANCES UNLESS SPECIFIED	 www.IDT.com	2975 Slender Way Santa Clara, CA 95054 PHONE: (408) 727-6116 FAX: (408) 492-8674	DECIMAL	APPROVALS	DATE	TITLE	SIZE	REV
ANGULAR					PG/PG PACKAGE OUTLINE (PG OR PA TOPMARK CODE) 4.4 mm BODY WIDTH ISSOP 65 mm PITCH	DRAWING NO.		
±					C	PSC-4056	05	
XXXXX					DO NOT SCALE DRAWING	SHEET 2 OF 3		

20-Lead TSSOP Package Outline and Package Dimensions, continued



## Ordering Information

Table 7. Ordering Information

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
8T33FS314PGGI	IDT8T33FS314PGGI	"Lead-Free" 20-Lead TSSOP	Tube	-40°C to 85°C
8T33FS314PGGI8	IDT8T33FS314PGGI	"Lead-Free" 20-Lead TSSOP	Tape & Reel	-40°C to 85°C
8T33FS314PYGI	IDT8T33FS314PYGI	"Lead-Free" 20-Lead SSOP	Tube	-40°C to 85°C
8T33FS314PYGI8	IDT8T33FS314PYGI	"Lead-Free" 20-Lead SSOP	Tape & Reel	-40°C to 85°C





## IMPORTANT NOTICE AND DISCLAIMER

RENESAS ELECTRONICS CORPORATION AND ITS SUBSIDIARIES ("RENESAS") PROVIDES TECHNICAL SPECIFICATIONS AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS OR IMPLIED, INCLUDING, WITHOUT LIMITATION, ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE, OR NON-INFRINGEMENT OF THIRD-PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for developers who are designing with Renesas products. You are solely responsible for (1) selecting the appropriate products for your application, (2) designing, validating, and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, or other requirements. These resources are subject to change without notice. Renesas grants you permission to use these resources only to develop an application that uses Renesas products. Other reproduction or use of these resources is strictly prohibited. No license is granted to any other Renesas intellectual property or to any third-party intellectual property. Renesas disclaims responsibility for, and you will fully indemnify Renesas and its representatives against, any claims, damages, costs, losses, or liabilities arising from your use of these resources. Renesas' products are provided only subject to Renesas' Terms and Conditions of Sale or other applicable terms agreed to in writing. No use of any Renesas resources expands or otherwise alters any applicable warranties or warranty disclaimers for these products.

(Disclaimer Rev.1.01)

### Corporate Headquarters

TOYOSU FORESIA, 3-2-24 Toyosu,  
Koto-ku, Tokyo 135-0061, Japan  
[www.renesas.com](http://www.renesas.com)

### Contact Information

For further information on a product, technology, the most up-to-date version of a document, or your nearest sales office, please visit [www.renesas.com/contact-us/](http://www.renesas.com/contact-us/).

### Trademarks

Renesas and the Renesas logo are trademarks of Renesas Electronics Corporation. All trademarks and registered trademarks are the property of their respective owners.