

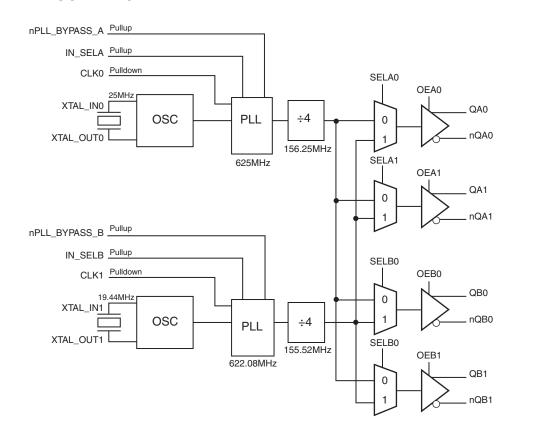
### **GENERAL DESCRIPTION**

**BLOCK DIAGRAM** 

The 843204I is a 4 output LVPECL Synthesizer optimized to generate Gigabit Ethernet and SONET reference clock frequencies and is a member of the family of high performance clock solutions from IDT. Using a 19.44MHz and 25MHz, 18pF parallel resonant crystal, 155.52MHz and 156.25MHz frequencies can be generated. The 843204I uses IDT's FemtoClock<sup>™</sup> low phase noise VCO technology and can achieve 1ps or lower typical RMS phase jitter. The 843204I is pack-aged in a 48-pin TSSOP package.

### **F**EATURES

- Four 3.3V LVPECL outputs
- Selectable crystal oscillator interface or LVCMOS/LVTTL single-ended input
- Supports the following output frequencies: 155.52MHz
  and 156.25MHz
- VCO range: 560MHz 680MHz
- RMS phase jitter @ 155.52MHz, using a 19.44MHz crystal (12kHz 20MHz): 0.98ps (typical)
- RMS phase jitter @ 156.25MHz, using a 19.44MHz crystal (1.875MHz 20MHz): 0.52ps (typical)
- Full 3.3V supply mode
- -40°C to 85°C ambient operating temperature
- Available in lead-free (RoHS 6) package



### **PIN ASSIGNMENT**

		/	1
nQA1 🗆	1	48	□ IN_SEL_A
QA1 🗆	2	47	CLK0
nQA0 🗆	3	46	☐ XTAL_IN0
QA0 🗆	4	45	□ XTAL_OUT0
nc 🗆	5	44	□ nc
Vcco_a 🗆	6	43	□ Vee
SELA1 🗆	7	42	□ OEA0
SELA0 🗆	8	41	OEA1
nPLL_BYPASS_A	9	40	□ Vcc
nc 🗆	10	39	□ Vcca_a
nc 🗆	11	38	🗆 nc
nc 🗆	12	37	🗆 nc
nc 🗆	13	36	☐ SELB0
XTAL_IN1 🗆	14	35	□ Vee
XTAL_OUT1 🗆	15	34	OEB0
CLK1 🗆	16	33	OEB1
IN_SEL_B	17	32	□ Vcc
nPLL_BYPASS_B	18	31	SELB1
Vссо_в 🗆	19	30	🗆 Vсса_в
nc 🗆	20	29	□ nc
QB0 🗆	21	28	🗆 nc
nQB0 🗆	22	27	🗆 nc
QB1 🗆	23	26	□ nc
nQB1 ⊡	24	25	□ nc
I			1

8432041 48 Lead TSSOP 6.1mm x 12.5mm x 0.925mm package body G Package Top View

#### TABLE 1. PIN DESCRIPTIONS

Number	Name	Туре		Description		
1, 2	nQA1, QA1	Output		Differential output pair. LVPECL interface levels.		
3, 4	nQA0, QA0	Output		Differential output pair. LVPECL interface levels.		
5, 10, 11, 12, 13, 20, 25, 26, 27, 28, 29, 37, 38, 44	nc	Unused		No connect.		
6	V <sub>cco_A</sub>	Power		Output supply pin for Bank A outputs.		
7	SELA1	Input	Pulldown	Select pin. When HIGH, selects QA1/nQA1 at 155.52MHz. When LOW, selects QA1/nQA1 at 156.25MHz. LVCMOS/LVTTL interface levels.		
8	SELA0	Input	Pulldown	Select pin. When HIGH, selects QA0/nQA0 at 155.52MHz. When LOW, selects QA1/nQA1 at 156.25MHz. LVCMOS/LVTTL interface levels.		
9	nPLL_BYPASS_A	Input	Pullup	When LOW, PLL is bypassed. When HIGH, PLL output is active. LVCMOS/LVTTL interface levels.		
14, 15	XTAL_IN1, XTAL_ OUT1	Input		Parallel resonant crystal interface. XTAL_OUT1 is the output, XTAL_IN1 is the input.		
16, 47	CLK1, CLK0	Input	Pulldown	LVCMOS/LVTTL clock inputs.		
21, 22	QB0, nQB0	Ouput		Differential output pair. LVPECL interface levels.		
17	IN_SEL_B	Input	Pullup	Select pin. When HIGH, selects XTAL1 inputs. When LOW, selects CLK1 input. LVCMOS/LVTTL interface levels.		
18	nPLL_BYPASS_B	Input	Pullup	When LOW, PLL is bypassed. When HIGH, PLL output is active. LVCMOS/LVTTL interface levels.		
19	V <sub>cco_B</sub>	Power		Output supply pin for Bank B outputs.		
23, 24	QB1, nQB1	Ouput		Differential output pair. LVPECL interface levels.		
31	SELB1	Input	Pullup	Select pin. When HIGH, selects QB1/nQB1 at 155.52MHz. When LOW, selects QB1/nQB1 at 156.25MHz. LVCMOS/LVTTL interface levels.		
30	V <sub>cca_b</sub>	Power		Analog supply pin for Bank B outputs.		
32, 40	V <sub>cc</sub>	Power		Core supply pins.		
33	OEB1	Input	Pullup	Output enable pin. When HIGH, QB1/nQB1 outputs are enable. LVCMOS/LVTTL interface levels.		
34	OEB0	Input	Pullup	Output enable pin. When HIGH, QB0/nQB0 outputs are enabled. LVCMOS/LVTTL interface levels.		
35, 43	V	Power		Negative supply pins.		
36	SELB0	Input	Pullup	Select pin. When HIGH, selects QB0/nQB0 at 155.52MHz. When LOW, selects QB0/nQB0 at 156.25MHz. LVCMOS/LVTTL interface levels.		
39	V <sub>cca_a</sub>	Power		Analog supply pin for Bank A outputs.		
41	OEA1	Input	Pullup	Output enable pin. When HIGH, QA1/nQA1 outputs are enabled. LVCMOS/LVTTL interface levels.		
42	OEA0	Input	Pullup	Output enable pin. When HIGH, QA0/nQA0 outputs are enabled. LVCMOS/LVTTL interface levels.		
45, 46	XTAL_OUT0, XTAL_IN0	Input		Parallel resonant crystal interface. XTAL_OUT0 is the output, XTAL_IN0 is the input.		
48	IN_SEL_A	Input	Pullup	Select pin. When HIGH, selects XTAL0 inputs. When LOW, selects CLK0 input. LVCMOS/LVTTL interface levels.		

NOTE: refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

#### TABLE 2. PIN CHARACTERISTICS

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C	Input Capacitance			4		pF
	Input Pulldown Resistor			51		kΩ
R	Input Pullup Resistor			51		kΩ

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#### ABSOLUTE MAXIMUM RATINGS

Supply Voltage, $V_{\rm cc}$	4.6V
Inputs, V	-0.5V to $V_{\rm cc}$ + 0.5V
Outputs, I	
Continuous Current	50mA
Surge Current	100mA
Package Thermal Impedance, $\boldsymbol{\theta}_{_{\!$	54.8°C/W (0 mps)
Storage Temperature, T <sub>stg</sub>	-65°C to 150°C

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

#### TABLE 3A. Power Supply DC Characteristics, $V_{cc} = V_{cco_A} = V_{cco_B} = 3.3V \pm 10\%$ , $V_{ee} = 0V$ , $T_{a} = -40^{\circ}C$ to $85^{\circ}C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>cc</sub>	Core Supply Voltage		2.97	3.3	3.63	V
V <sub>cca_a</sub> + V <sub>cca_b</sub>	Analog Supply Voltage		V <sub>cc</sub> – 0.22	3.3	V <sub>cc</sub>	V
V V <sub>cco_B</sub>	Output Supply Voltage		2.97	3.3	3.63	V
I	Power Supply Current				170	mA
I <sub>cc</sub>	Core Supply Current				125	mA
$I_{CCA_A} + I_{CCA_B}$	Analog Supply Current				22	mA
I <sub>cco_A</sub> +I <sub>cco_B</sub>	Output Supply Current				23	mA

#### TABLE 3B. LVCMOS / LVTTL DC CHARACTERISTICS, $V_{cc} = V_{cco,a} = V_{cco,b} = 3.3V \pm 10\%$ , $V_{ee} = 0V$ , $T_{a} = -40^{\circ}$ C to $85^{\circ}$ C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V	Input High Vol	tage		2		V <sub>cc</sub> + 0.3	V
V	Input Low Voltage			-0.3		0.8	V
Input		CLK0, CLK1, SELA0, SELA1	$V_{\rm cc} = V_{\rm in} = 3.63 V$			150	μA
	Input High Current	nPLL_BYPASS_A, nPLL_ BYPASS_B, IN_SEL_A, IN_SEL_B, SELB1, SELB0, OEB0, OEB1, OEA0, OEA1	$V_{\rm cc} = V_{\rm in} = 3.63 V$			5	μΑ
		CLK0, CLK1, SELA0, SELA1	$V_{_{\rm CC}} = 3.63 \text{V},  \text{V}_{_{\rm IN}} = 0 \text{V}$	-5			μA
I <sub>n</sub>	Input Low Current	nPLL_BYPASS_A, nPLL_ BYPASS_B, IN_SEL_A, IN_SEL_B, SELB1, SELB0, OEB0, OEB1, OEA0, OEA1	$V_{_{CC}} = 3.63V, V_{_{IN}} = 0V$	-150			μΑ

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OH</sub>	Output High Voltage; NOTE 1		V <sub>cco</sub> - 1.4		V <sub>cco</sub> - 0.9	V
V	Output Low Voltage; NOTE 1		V <sub>cco</sub> - 2.0		V <sub>cco</sub> - 1.7	V
	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs terminated with 50  $\!\Omega$  to V  $_{_{\rm CCO}}$  - 2V.

#### TABLE 4. CRYSTAL CHARACTERISTICS

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation		Fundamental			
Frequency		19.44		25	MHz
Equivalent Series Resistance (ESR)				50	Ω
Shunt Capacitance				7	pF

NOTE: Characterized using an 18pF parallel resonant crystal.

### Table 5. AC Characteristics, $V_{cc} = V_{cco_{-B}} = 3.3V \pm 10\%$ , $V_{ee} = 0V$ , $T_{A} = -40^{\circ}C$ to $85^{\circ}C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
tsk(b) B		SELB0 = 1; OEB0 = 1		155.52		MHz
	Output Frequency	SELA0 = 0; OEA0 = 1		156.25		MHz
tsk(b)	Bank Skew; NOTE 1, 2				50	ps
+;;;+(0)	RMS Phase Jitter (Random);	155.52MHz, (12kHz - 20MHz)		0.98		ps
(jii(Ø)	NOTE 3	156.25MHz, (1.875MHz - 20MHz)		0.52		ps
t <sub>R</sub> / t <sub>F</sub>	Output Rise/Fall Time	20% to 80%	250		600	ps
odc	Output Duty Cycle		47		53	%

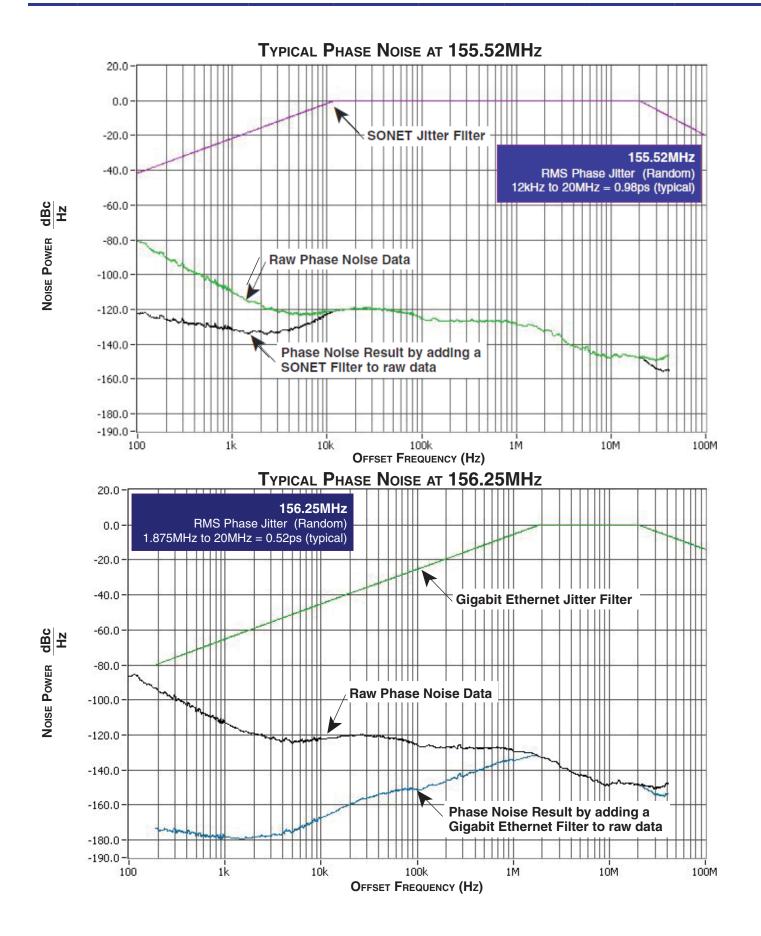
NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: Defined as skew within a bank of outputs at the same supply voltage and with equal load conditions.

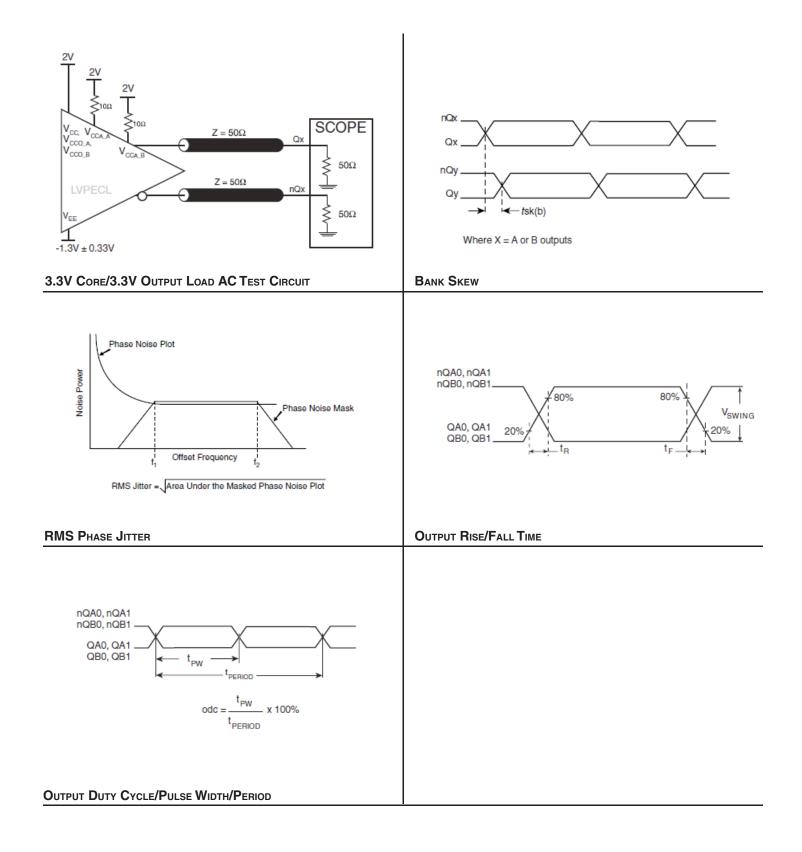
NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 3: See Phase Noise plot.

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# **PARAMETER MEASUREMENT INFORMATION**



## **APPLICATION** INFORMATION

#### Power Supply Filtering Techniques

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The 8432041 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL.  $V_{cc}$ ,  $V_{ccA,A}$ ,  $V_{ccA,B}$ ,  $V_{ccA,B}$ , and  $V_{cco,B}$  should be individually connected to the power supply plane through vias, and 0.01µF bypass capacitors should be used for each pin. *Figure 1* illustrates this for a generic  $V_{cc}$  pin and also shows that  $V_{ccA}$  requires that an additional10 $\Omega$  resistor along with a 10µF bypass capacitor be connected to the  $V_{ccA}$  pin.

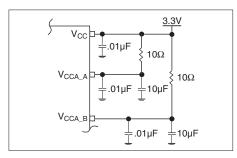


FIGURE 1. POWER SUPPLY FILTERING

#### **RECOMMENDATIONS FOR UNUSED INPUT AND OUTPUT PINS**

#### **INPUTS:**

#### **C**RYSTAL INPUTS

For applications not requiring the use of the crystal oscillator input, both XTAL\_IN and XTAL\_OUT can be left floating. Though not required, but for additional protection, a  $1k\Omega$  resistor can be tied from XTAL\_IN to ground.

#### **CLK** INPUTS

For applications not requiring the use of a clock input, it can be left floating. Though not required, but for additional protection, a  $1k\Omega$  resistor can be tied from the CLK input to ground.

#### LVCMOS CONTROL PINS

All control pins have internal pullups or pulldowns; additional resistance is not required but can be added for additional protection. A  $1k\Omega$  resistor can be used.

#### **O**UTPUTS:

#### LVPECL OUTPUTS

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

#### **C**RYSTAL INPUT INTERFACE

The 8432041 has been characterized with 18pF parallel resonant crystals. The capacitor values shown in *Figure 2* below

were determined using an 18pF parallel resonant crystal and were chosen to minimize the ppm error.

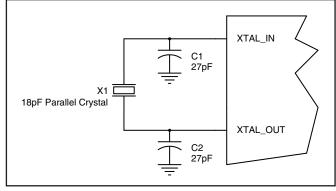


FIGURE 2. CRYSTAL INPUT INTERFACE

#### LVCMOS TO XTAL INTERFACE

The XTAL\_IN input can accept a single-ended LVCMOS signal through an AC coupling capacitor. A general interface diagram is shown in *Figure 3*. The XTAL\_OUT pin can be left floating. The input edge rate can be as slow as 10ns. For LVCMOS signals, it is recommended that the amplitude be reduced from full swing to half swing in order to prevent signal interference with the power rail and to reduce noise. This configuration requires that the output

impedance of the driver (Ro) plus the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This can be done in one of two ways. First, R1 and R2 in parallel should equal the transmission line impedance. For most 50 $\Omega$  applications, R1 and R2 can be 100 $\Omega$ . This can also be accomplished by removing R1 and making R2 50 $\Omega$ .

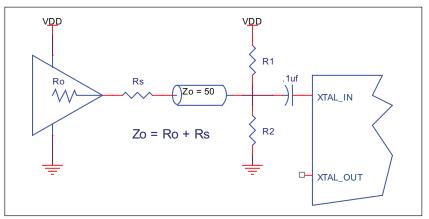


FIGURE 3. GENERAL DIAGRAM FOR LVCMOS DRIVER TO XTAL INPUT INTERFACE

#### **TERMINATION FOR 3.3V LVPECL OUTPUTS**

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

FOUT and nFOUT are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive  $50\Omega$  transmission

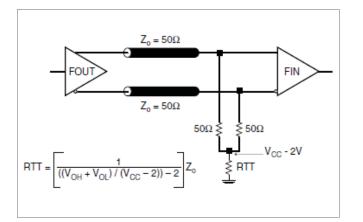


FIGURE 4A. LVPECL OUTPUT TERMINATION

lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

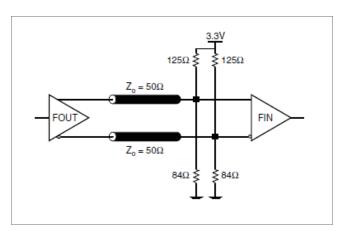


FIGURE 4B. LVPECL OUTPUT TERMINATION

## **Power Considerations**

This section provides information on power dissipation and junction temperature for the 843204I. Equations and example calculations are also provided.

#### 1. Power Dissipation.

The total power dissipation for the 843204I is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for  $V_{cc} = 3.3V + 10\% = 3.63V$ , which gives worst case results. **NOTE:** Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)<sub>MAX</sub> = V<sub>CC\_MAX</sub> \* I<sub>EE\_MAX</sub> = 3.63V \* 170mA = 617.10mW
- Power (outputs)<sub>MAX</sub> = 30mW/Loaded Output pair
  If all outputs are loaded, the total power is 4 \* 30mW = 120mW

**Total Power**  $_{Max}$  (3.63V, with all outputs switching) = 617.10mW + 120mW = **737.10mW** 

#### 2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS<sup>™</sup> devices is 125°C.

The equation for Tj is as follows: Tj =  $\theta_{JA} * Pd_{total} + T_A$ 

Tj = Junction Temperature

 $\theta_{\text{JA}}$  = Junction-to-Ambient Thermal Resistance

Pd\_total = Total Device Power Dissipation (example calculation is in section 1 above)

T<sub>A</sub> = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming a moderate air flow of 1 meter per second and a multi-layer board, the appropriate value is 51°C/W per Table 6 below.

Therefore, Tj for an ambient temperature of  $85^{\circ}$ C with all outputs switching is:  $85^{\circ}$ C + 0.737W \*  $51^{\circ}$ C/W = 122.6°C. This is below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

#### TABLE 6. THERMAL RESISTANCE $\theta_{JA}$ FOR 48-PIN TSSOP, FORCED CONVECTION

θ <sub>JA</sub> by Velocity (Meters Per Second)					
Multi-Layer PCB, JEDEC Standard Te	st Boards	<b>0</b> 54.8°C/W	<b>1</b> 51.0°C/W	<b>2.5</b> 49.1°C/W	

#### 3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in Figure 5.

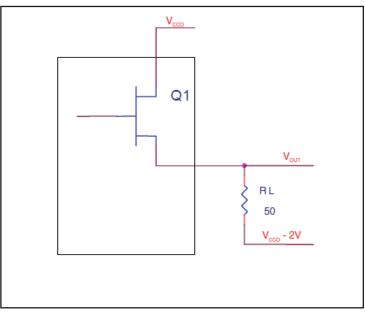


FIGURE 5. LVPECL DRIVER CIRCUIT AND TERMINATION

To calculate worst case power dissipation into the load, use the following equations which assume a 50 $\Omega$  load, and a termination voltage of V<sub>cco</sub> – 2V.

• For logic high,  $V_{OUT} = V_{OH, MAX} = V_{CCO, MAX} - 0.9V$ 

$$(V_{\text{CCO}_{MAX}} - V_{\text{OH}_{MAX}}) = 0.9V$$

• For logic low,  $V_{OUT} = V_{OL_{MAX}} = V_{CCO_{MAX}} - 1.7V$ 

$$(V_{\text{CCO}_{\text{MAX}}} - V_{\text{OL}_{\text{MAX}}}) = 1.7V$$

Pd\_H is power dissipation when the output drives high. Pd\_L is the power dissipation when the output drives low.

 $\mathsf{Pd}_{-}\mathsf{H} = [(\mathsf{V}_{\mathsf{OH}\_MAX} - (\mathsf{V}_{\mathsf{CCO}\_MAX} - 2\mathsf{V}))/\mathsf{R}_{\text{L}}] * (\mathsf{V}_{\mathsf{CCO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) = [(2\mathsf{V} - (\mathsf{V}_{\mathsf{CCO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}))/\mathsf{R}_{\text{L}}] * (\mathsf{V}_{\mathsf{CCO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) = [(2\mathsf{V} - (\mathsf{V}_{\mathsf{CCO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}))/\mathsf{R}_{\text{L}}] * (\mathsf{V}_{\mathsf{CCO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) = [(2\mathsf{V} - (\mathsf{V}_{\mathsf{CCO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}))/\mathsf{R}_{\text{L}}] * (\mathsf{V}_{\mathsf{CCO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) = [(\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) + (\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}))/\mathsf{R}_{\text{L}}] * (\mathsf{V}_{\mathsf{CCO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) = [(\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) + (\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}))/\mathsf{R}_{\text{L}}] * (\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) = [(\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) + (\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}))/\mathsf{R}_{\text{L}}] * (\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) = [\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX})] = [\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) = [\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX})] = [\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX}) = [\mathsf{V}_{\mathsf{CO}\_MAX} - \mathsf{V}_{\mathsf{OH}\_MAX})]$ 

 $[(2V - 0.9V)/50\Omega] * 0.9V = 19.8mW$ 

 $Pd_{L} = [(V_{\text{ol_max}} - (V_{\text{cco_max}} - 2V))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{L}}] * (V_{\text{cco_max}} - V_{\text{ol_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}}))/R_{\text{cco_max}}) = [(2V - (V_{\text{cco_max}} - V_{\text{ol_max}})]$ 

 $[(2V - 1.7V)/50\Omega] * 1.7V = 10.2mW$ 

Total Power Dissipation per output pair =  $Pd_H + Pd_L = 30mW$ 

# **R**ELIABILITY INFORMATION

# TABLE 7. $\boldsymbol{\theta}_{_{JA}} \text{vs.}$ Air Flow Table for 48 Lead TSSOP

$ heta_{JA}$ by Veloci	ty (Meters Per	Second)		
Multi-Layer PCB, JEDEC Standard Test Boards	<b>0</b> 54.8°C/W	<b>1</b> 51.0°C/W	<b>2.5</b> 49.1°C/W	

#### TRANSISTOR COUNT

The transistor count for 843204I is: 4090

## RENESAS

PACKAGE OUTLINE - G SUFFIX FOR 48 LEAD TSSOP

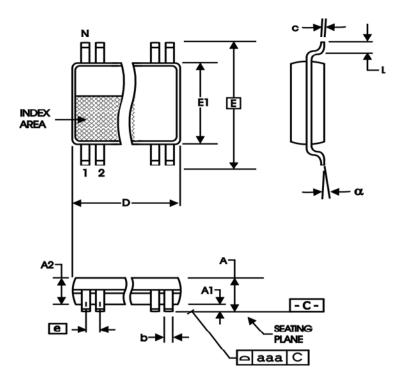


TABLE 8. PACKAGE DIMENSIONS

SYMBOL	Millimeters		
STMBOL	Minimum	Maximum	
N	48		
A		1.20	
A1	0.05	0.15	
A2	0.80	1.05	
b	0.17	0.27	
с	0.09	0.20	
D	12.40	12.60	
E	8.10 BASIC		
E1	6.00	6.20	
е	0.50 BASIC		
L	0.45	0.75	
α	0°	8°	
aaa		0.10	

Reference Document: JEDEC Publication 95, MO-153

#### TABLE 9. ORDERING INFORMATION

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
843204AGILF	ICS843204AGILF	48 Lead "Lead-Free" TSSOP	tube	-40°C to 85°C
843204AGILFT	ICS843204AGILF	48 Lead "Lead-Free" TSSOP	tape & reel	-40°C to 85°C

	REVISION HISTORY SHEET					
Rev	Table	Page	Description of Change			
A	Т9	8 14	Crystal Input Interface, Figure 2, changed Cq/C2 to 27pF. Ordering Information Table - deleted "ICS" prefix from part/order number column.	3/18/09		
А	Т9	14	Ordering Information - removed leaded devices. Updated data sheet format.	10/16/15		
A	Т9	1 14	General Description - Removed Hiperclocks. Ordering Information - Removed quantity in tape and reel. Deleted LF note below table. Updated Header and Footer.	1/18/16		



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TOYOSU FORESIA, 3-2-24 Toyosu, Koto-ku, Tokyo 135-0061, Japan www.renesas.com

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